

# Design IV

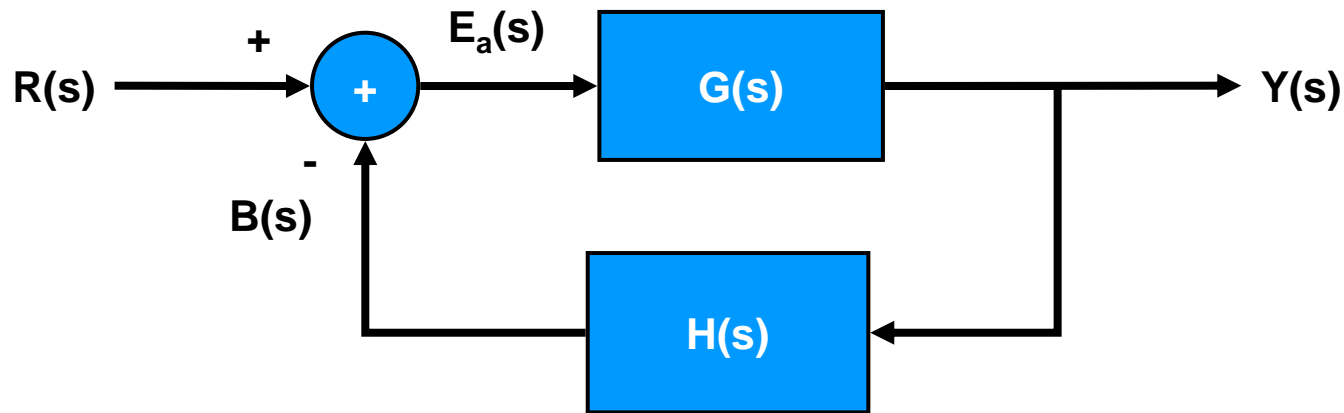
## E232 Fall 07

Class 14

Bruce McNair  
bmcnair@stevens.edu

# Transfer Function of a Feedback System

- Consider a generic feedback control system



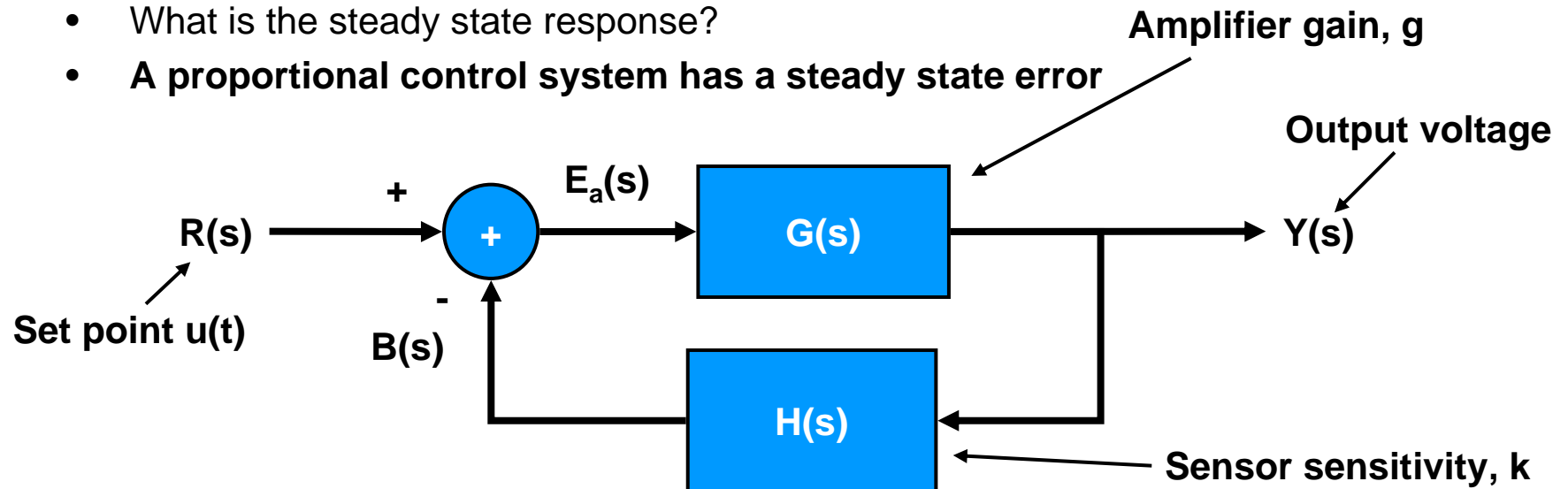
$$Y(s) = (R(s) - Y(s)H(s))G(s)$$

$$Y(s)[1 + G(s)H(s)] = R(s)G(s)$$

$$\frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$

# Steady State Response

- Consider the hot-wire anemometer as a feedback control system
- What is the steady state response?
- **A proportional control system has a steady state error**



$$Y(s) = \frac{g}{1 + gk} \frac{1}{s}$$

**From the final value theorem:**

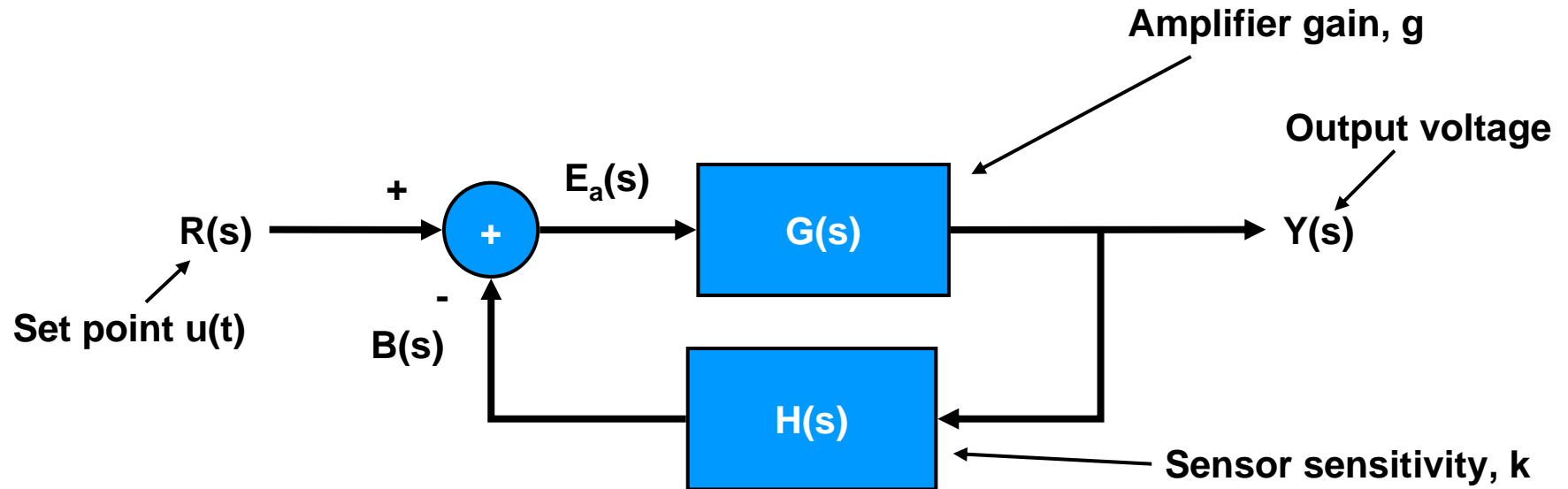
$$\lim_{t \rightarrow \infty} z(t) = \lim_{s \rightarrow 0} sZ(s)$$

$$\lim_{t \rightarrow \infty} e_a(t) = \lim_{s \rightarrow 0} s \frac{1}{s(1 + gk)}$$

$$\lim_{t \rightarrow \infty} e_a(t) = \frac{1}{1 + gk}$$

# Steady State Response

- To reduce the steady state error:



But, what if error signal was something like:

$$\lim_{t \rightarrow \infty} e_a(t) = \lim_{s \rightarrow 0} s \frac{1}{s \left( 1 + g \frac{k}{s} \right)}$$

Then:

$$\lim_{t \rightarrow \infty} e_a(t) = \lim_{s \rightarrow 0} \frac{1}{\left( 1 + g \frac{k}{s} \right)} = \frac{s}{s + gk} = 0$$

Remember

$$\frac{1}{s} \equiv \int dt$$

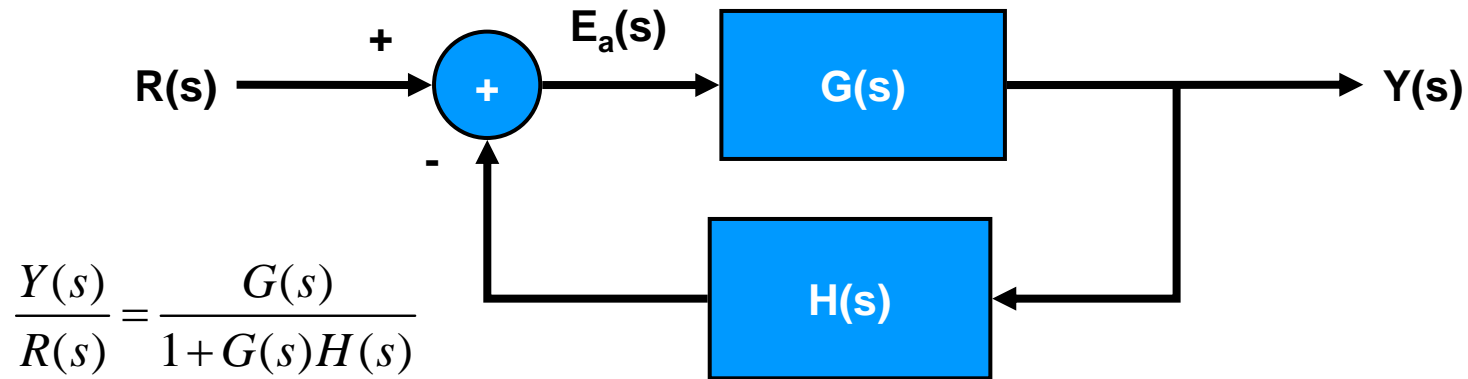
Integration of error signal allows steady state error to go to zero

# Today's topics

- Control systems
  - Unity feedback
  - Transient response
  - PID control
  - System types
  - Steady state error
  - Performance indices

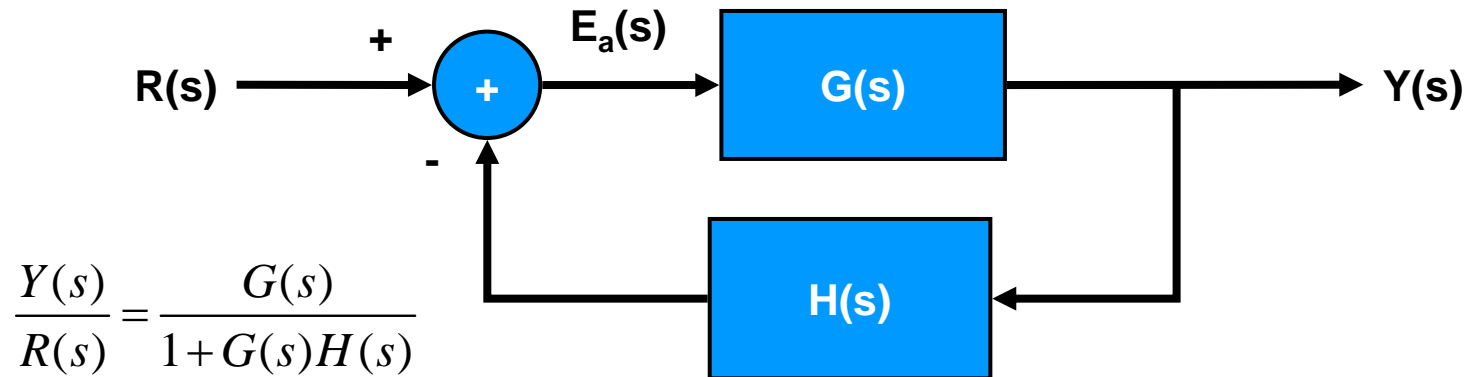
# Unity Feedback

- Generic control system

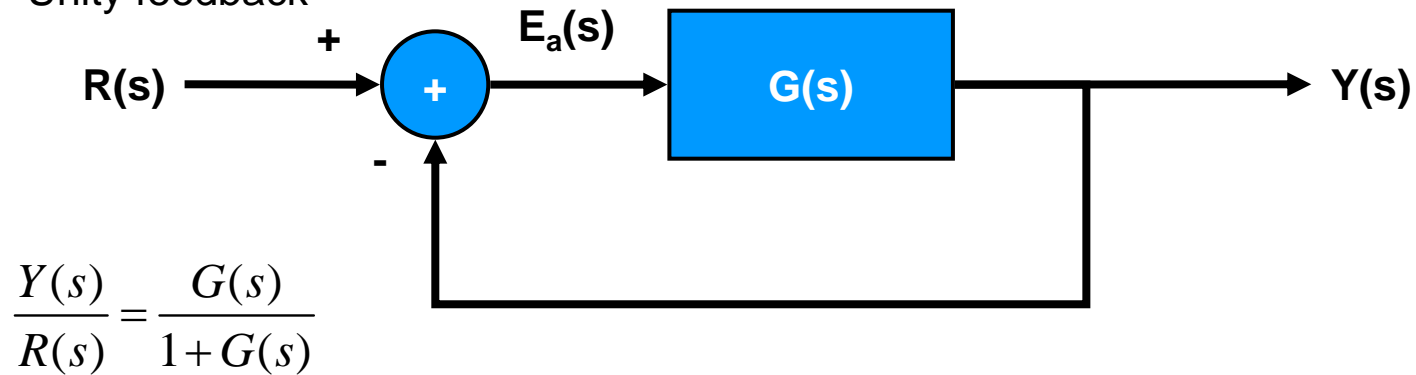


# Unity Feedback

- Generic control system

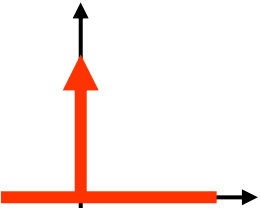
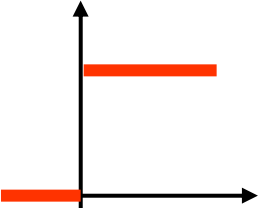
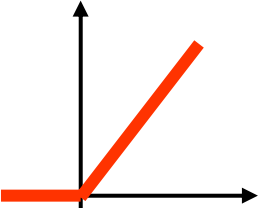
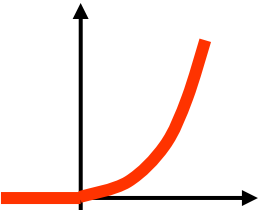


- Unity feedback



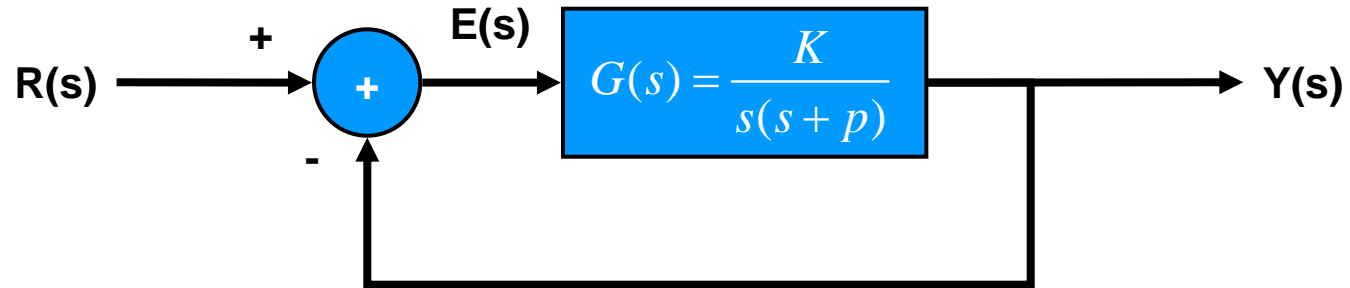
# Transient Response

- Test signals

	$r(t)$	$R(s)$
	$r_{\varepsilon}(t) = \begin{cases} \frac{1}{\varepsilon} & -\frac{\varepsilon}{2} \leq t \leq \frac{\varepsilon}{2} \\ 0 & \text{otherwise} \end{cases}$	1
	$r(t) = \begin{cases} A & t > 0 \\ 0 & t < 0 \end{cases}$	$\frac{A}{s}$
	$r(t) = \begin{cases} At & t > 0 \\ 0 & t < 0 \end{cases}$	$\frac{A}{s^2}$
	$r(t) = \begin{cases} At^2 & t > 0 \\ 0 & t < 0 \end{cases}$	$\frac{2A}{s^3}$

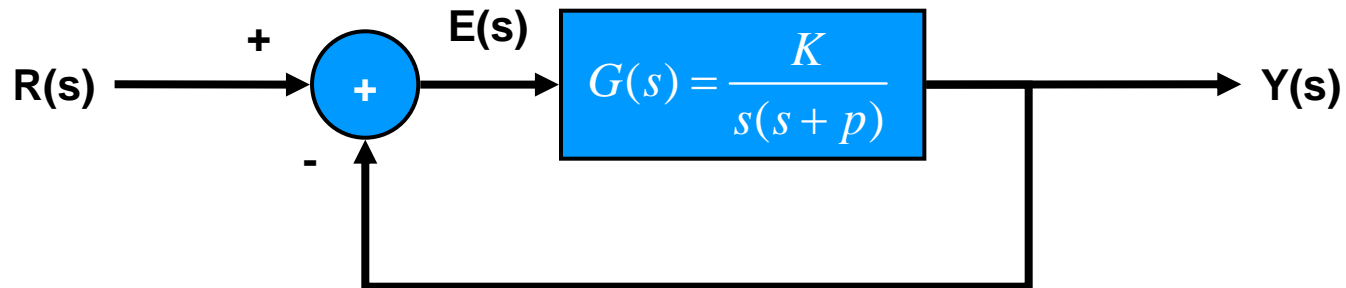
# Second Order System Performance

- Consider the simple control system:



# Second Order System Performance

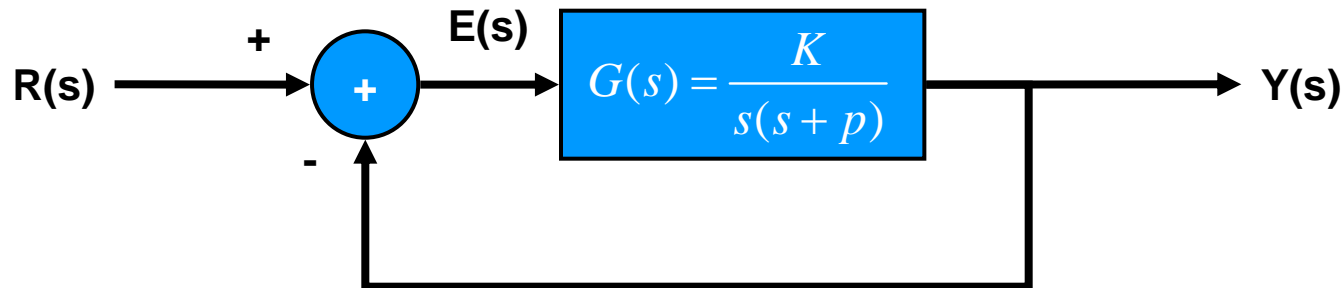
- Consider the simple control system:



$$Y(s) = \frac{G(s)}{1 + G(s)} R(s) = \frac{K}{s^2 + ps + K} R(s)$$

# Second Order System Performance

- Consider the simple control system:

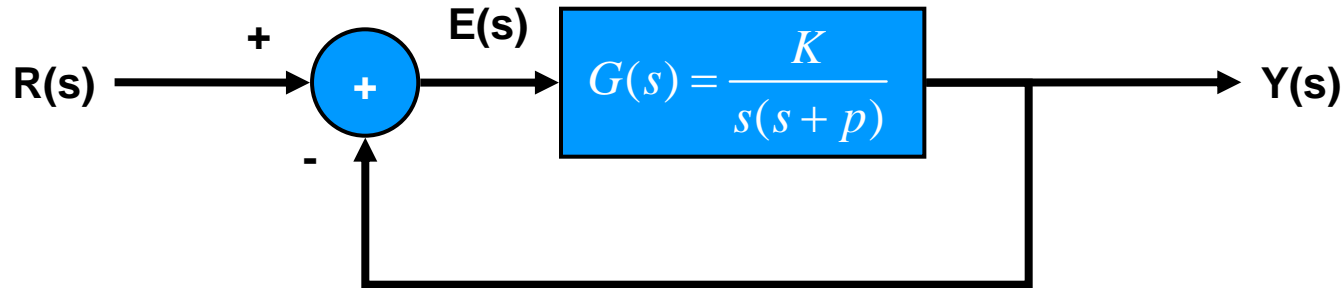


$$Y(s) = \frac{G(s)}{1+G(s)} R(s) = \frac{K}{s^2 + ps + K} R(s)$$

$$Y(s) = \frac{\omega_n^2}{(s^2 + 2\zeta\omega_n s + \omega_n^2)} R(s)$$

# Second Order System Performance

- Consider the simple control system:



$$Y(s) = \frac{G(s)}{1+G(s)} R(s) = \frac{K}{s^2 + ps + K} R(s)$$

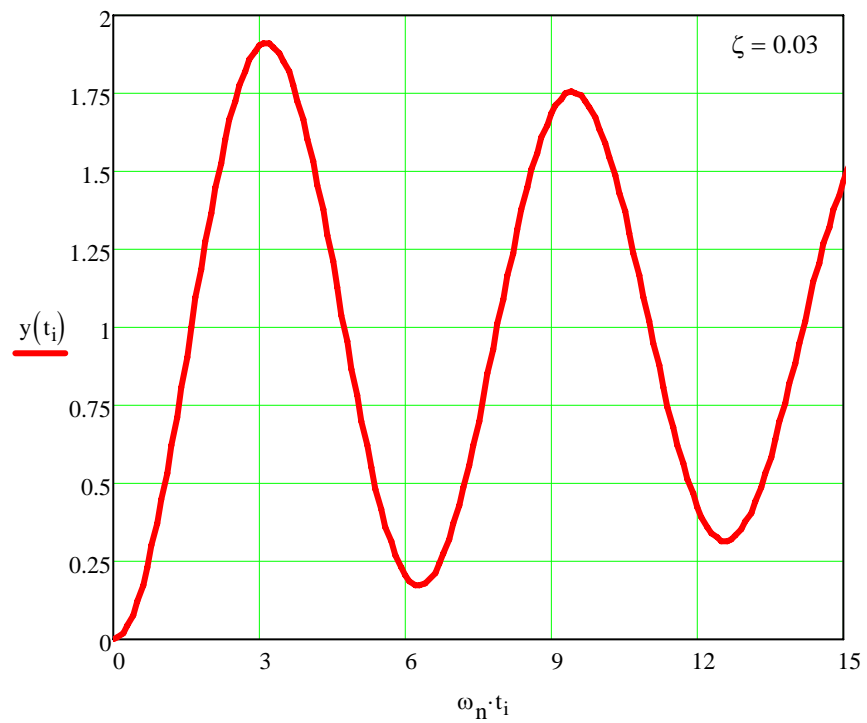
$$Y(s) = \frac{\omega_n^2}{(s^2 + 2\zeta\omega_n s + \omega_n^2)} R(s)$$

With  $r(t) = u(t)$ :

$$y(t) = 1 - \frac{1}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin\left(\omega_n \sqrt{1-\zeta^2} t + \cos^{-1}(\zeta)\right)$$

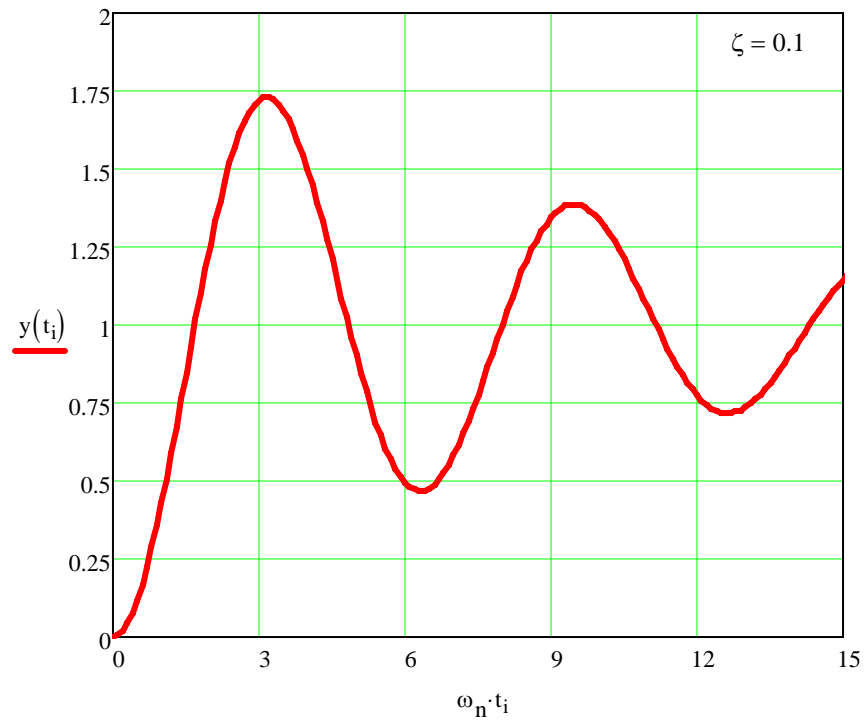
# Second Order System Performance

- Effect of damping factor,  $\zeta$



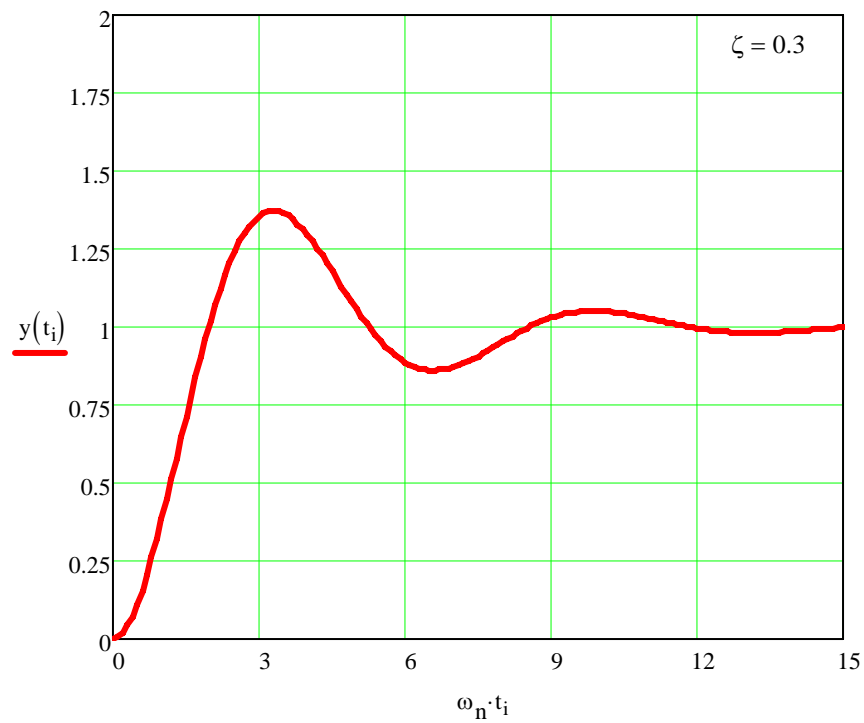
# Second Order System Performance

- Effect of damping factor,  $\zeta$



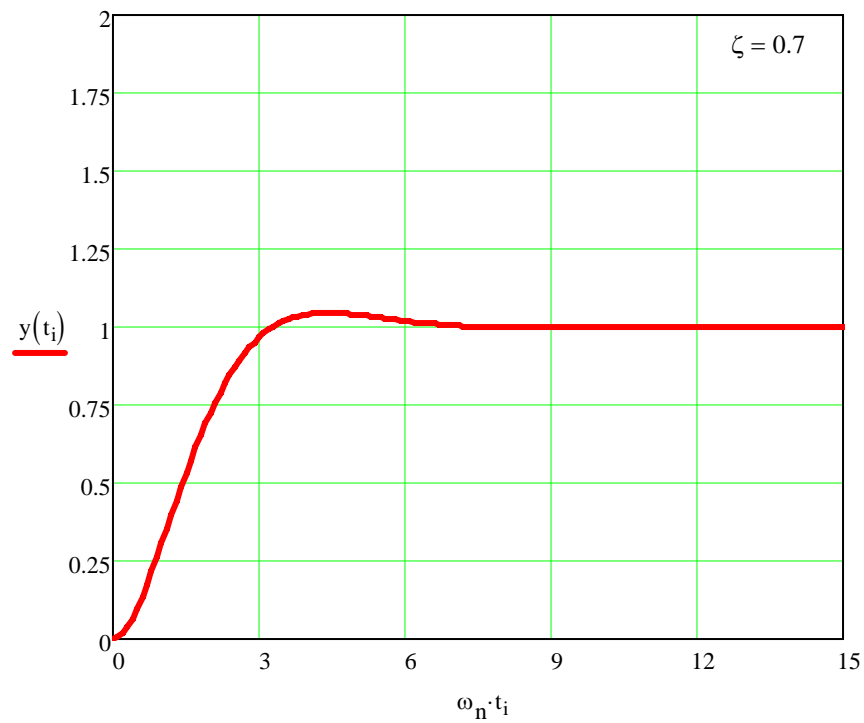
# Second Order System Performance

- Effect of damping factor,  $\zeta$



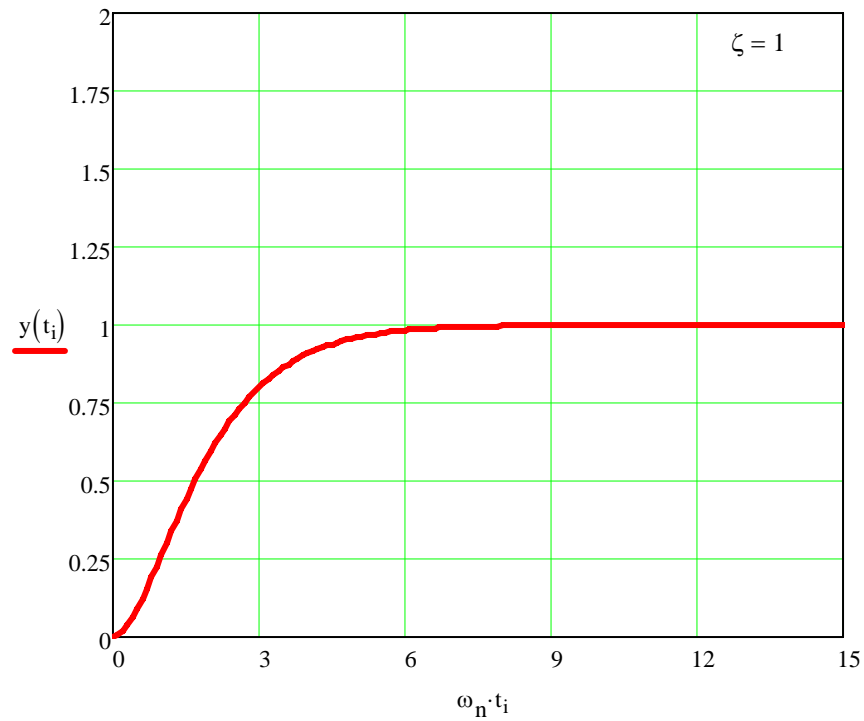
# Second Order System Performance

- Effect of damping factor,  $\zeta$



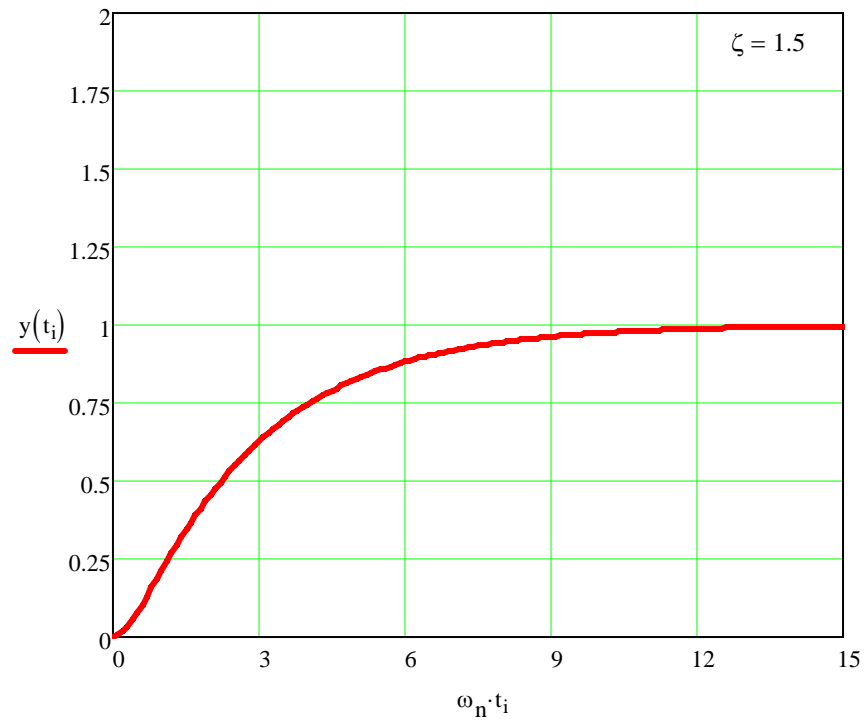
# Second Order System Performance

- Effect of damping factor,  $\zeta$



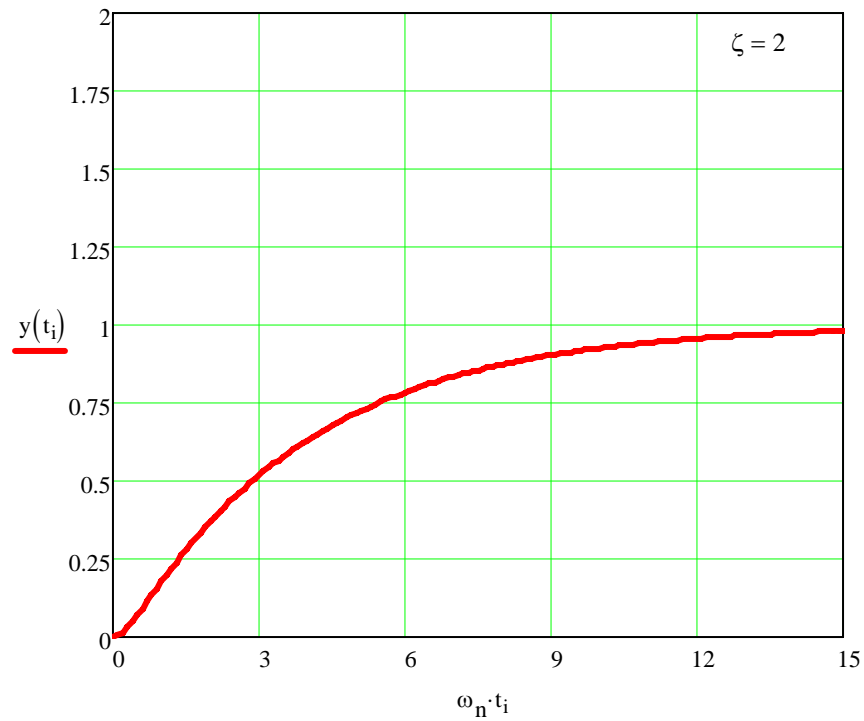
# Second Order System Performance

- Effect of damping factor,  $\zeta$



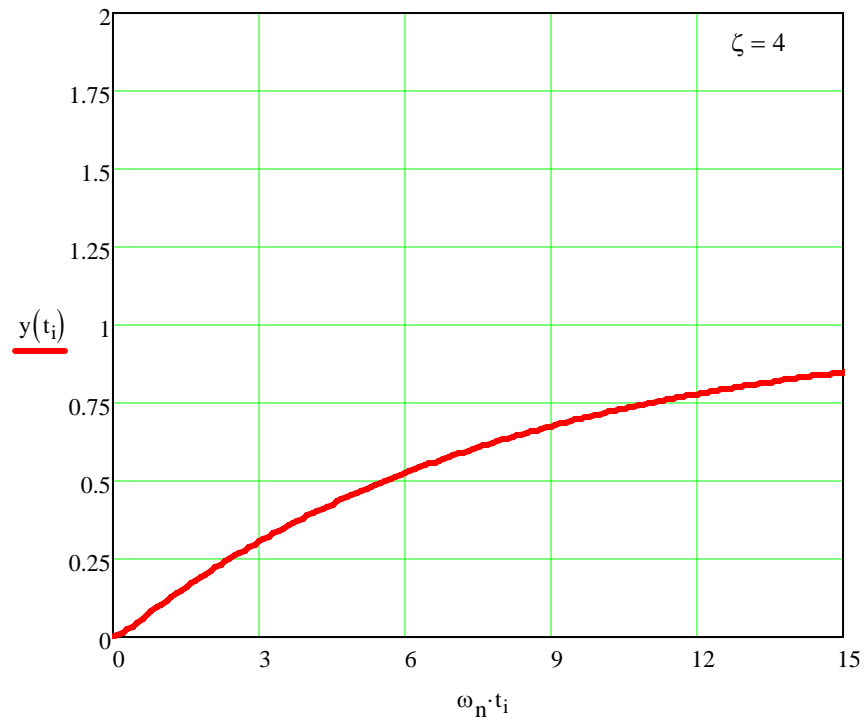
# Second Order System Performance

- Effect of damping factor,  $\zeta$



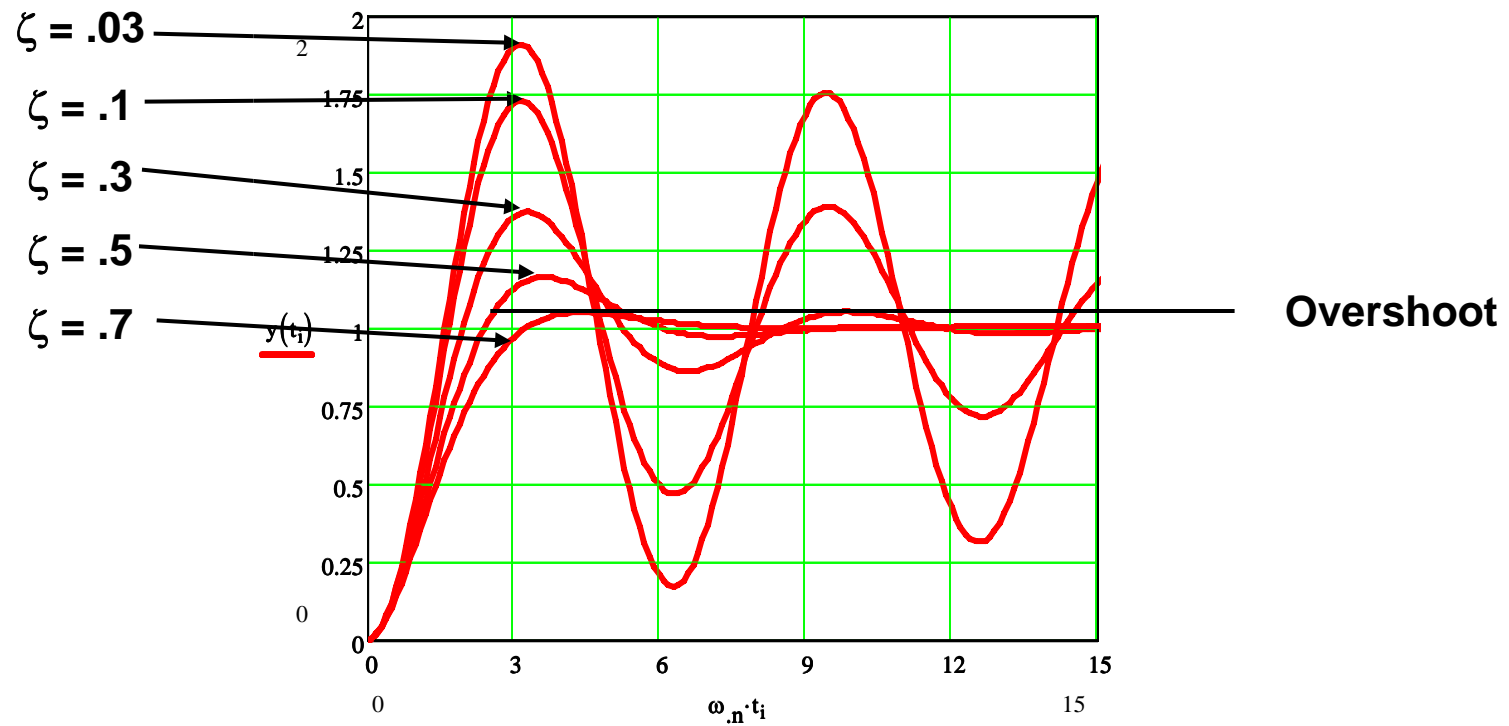
# Second Order System Performance

- Effect of damping factor,  $\zeta$



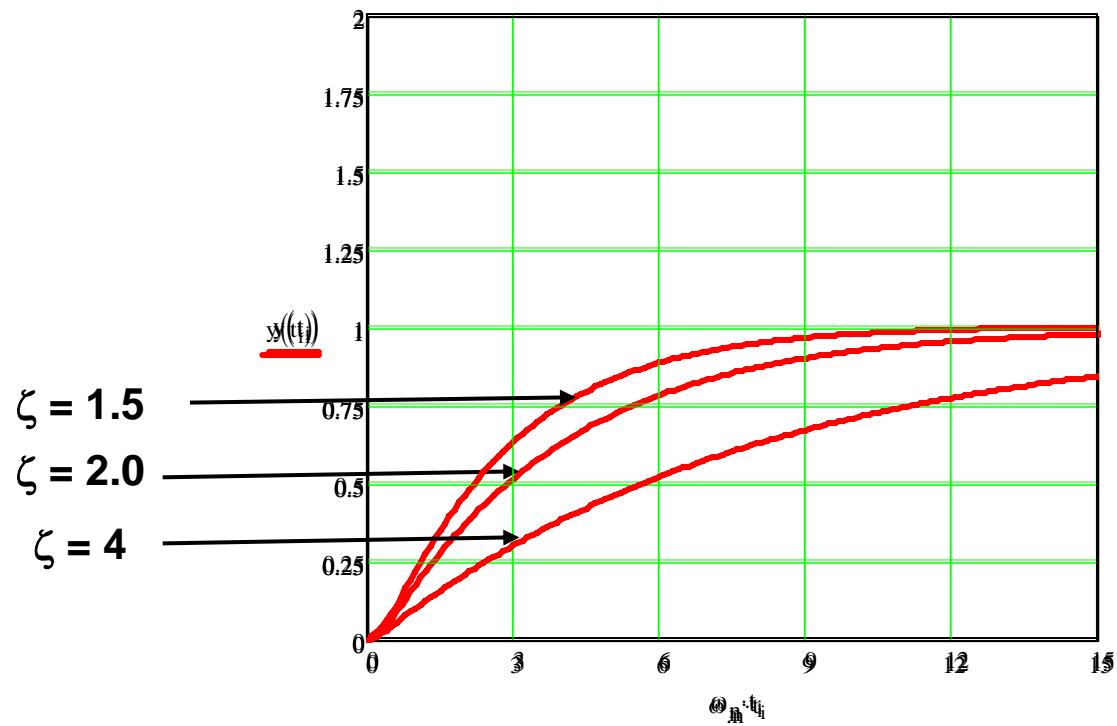
# Second Order System Performance

- Under damped,  $\zeta < 1$



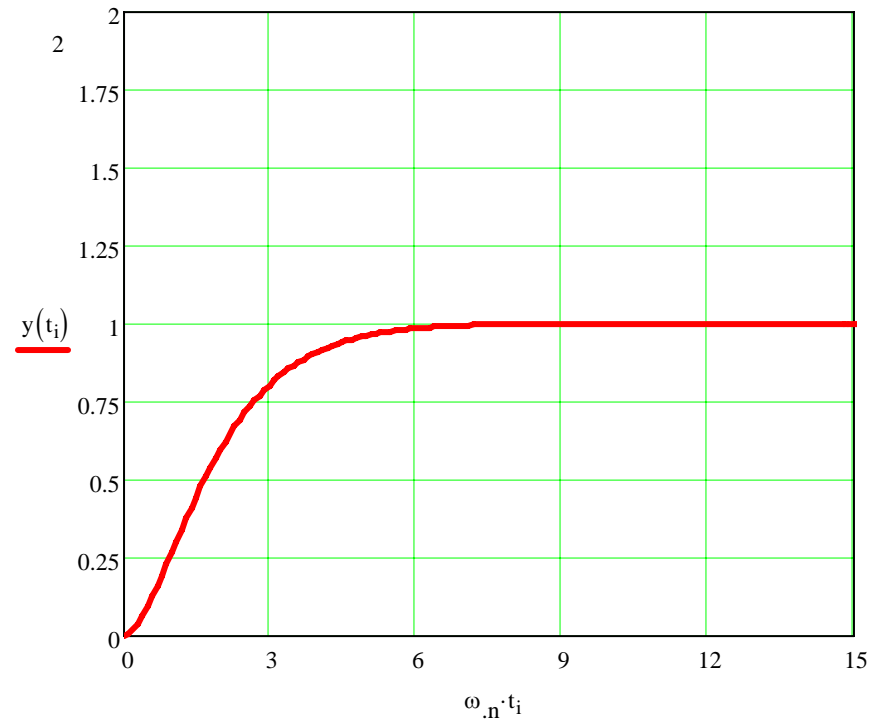
# Second Order System Performance

- Over damped,  $\zeta > 1$



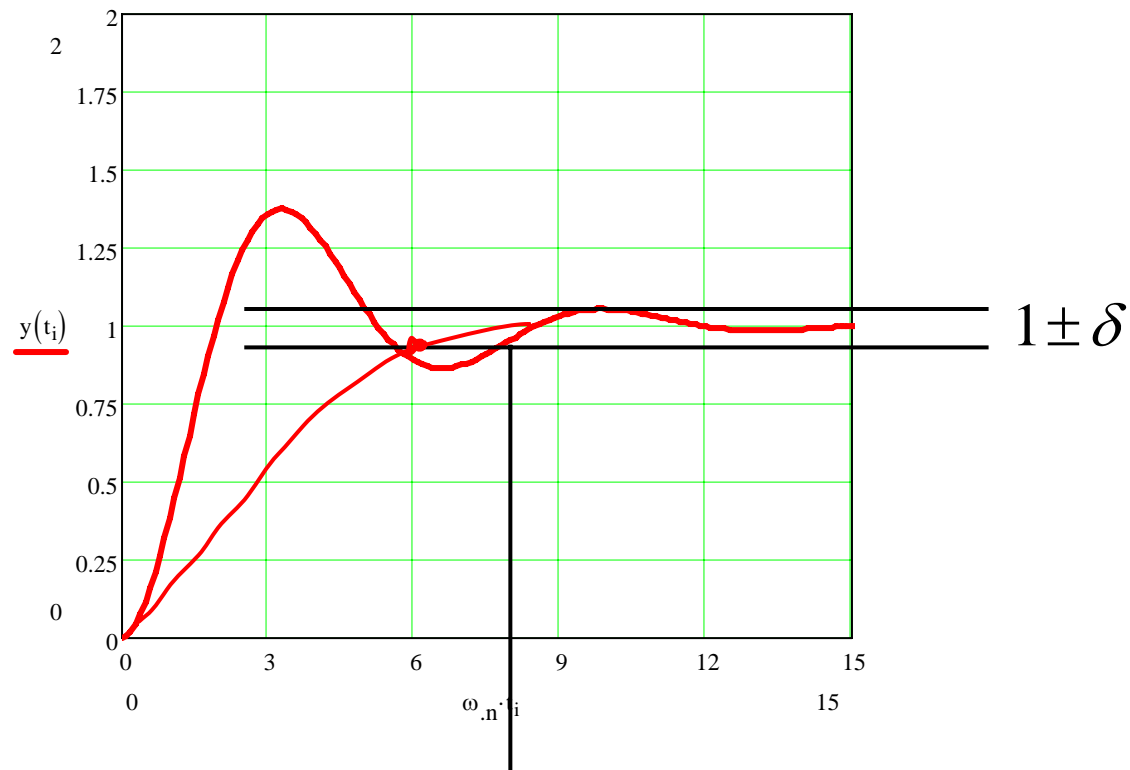
# Second Order System Performance

- Critically damped,  $\zeta = 1$



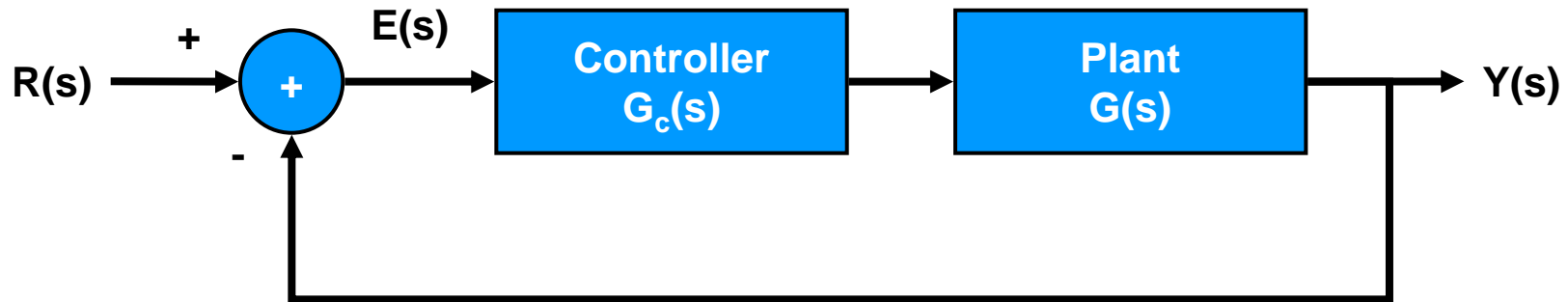
# Second Order System Performance

$$y(t) = 1 - \frac{1}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin\left(\omega_n \sqrt{1-\zeta^2} t + \cos^{-1}(\zeta)\right)$$



**Setting time**  $e^{-\zeta\omega_n T_s} < \delta$

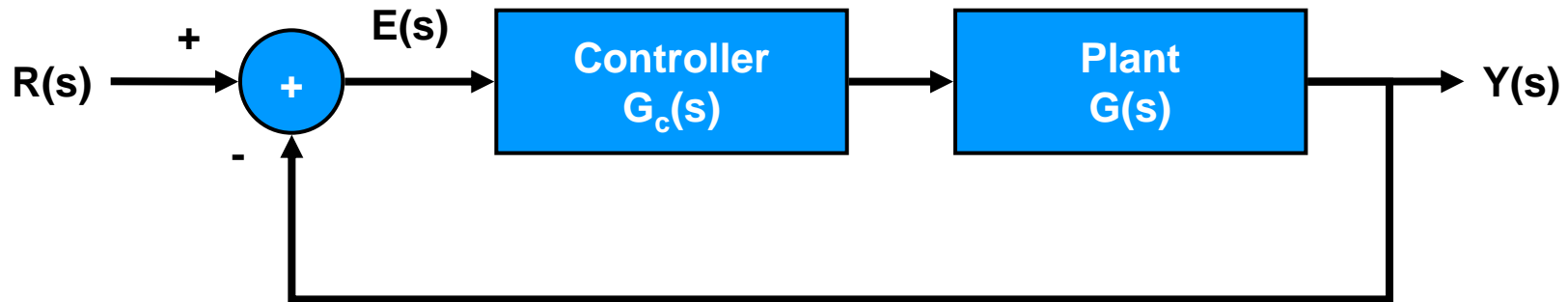
# Proportional, Integral, Derivative (PID) Control



$$G_c(s) = K_P + \frac{K_I}{s} + \frac{K_D s}{\tau_d s + 1}$$

$$G_c(s) \approx K_P + \frac{K_I}{s} + K_D s$$

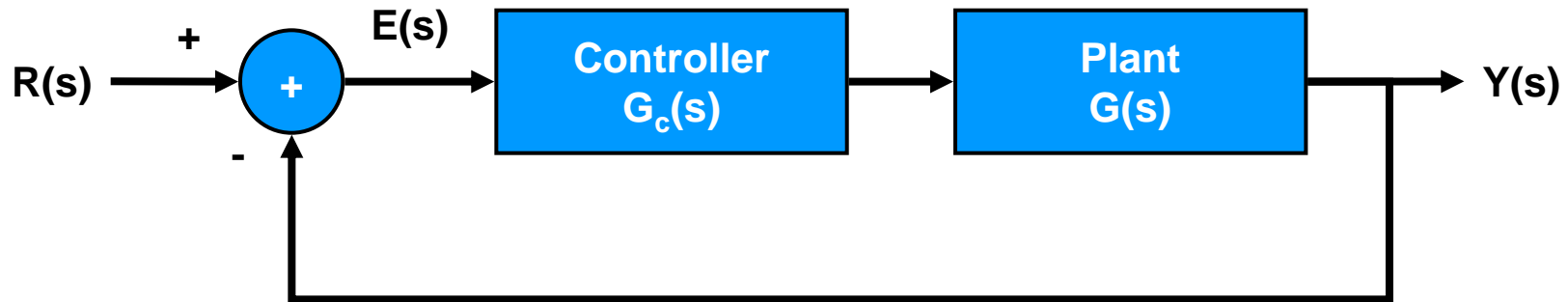
# Proportional Plus Integral (PI) Control



$$G_c(s) = K_P + \frac{K_I}{s} + \frac{K_D s}{\tau_d s + 1}$$

$$G_c(s) \approx K_P + \frac{K_I}{s} + \cancel{K_D s}$$

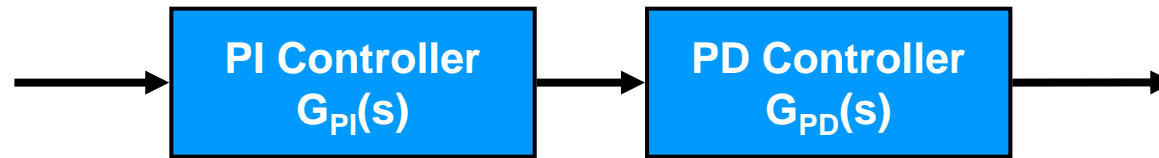
# Proportional Plus Derivative (PD) Control



$$G_c(s) = K_P + \frac{K_I}{s} + \frac{K_D s}{\tau_d s + 1}$$

$$G_c(s) \approx K_P + \cancel{\frac{K_I}{s}} + K_D s$$

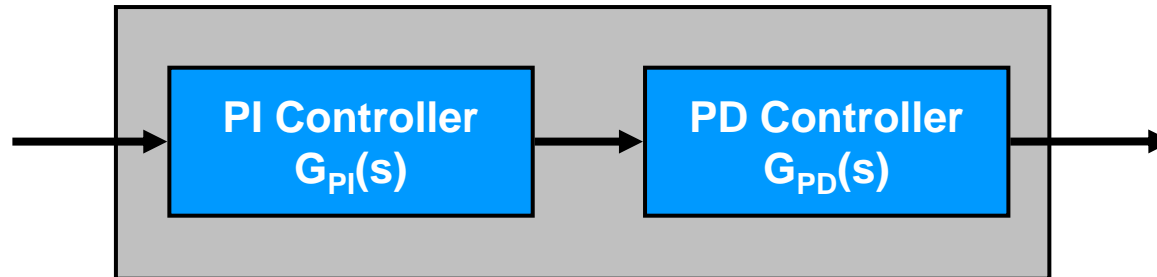
# PID Control = PD + PI



$$G_{PI}(s) = K'_P + \frac{K'_I}{s}$$

$$G_{PD}(s) = K''_P + K''_D s$$

# PID Control = PD + PI

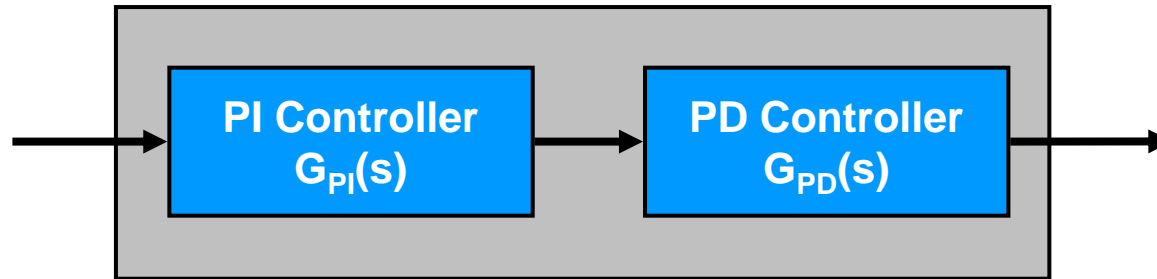


$$G_{PI}(s) = K'_P + \frac{K'_I}{s}$$

$$G_{PD}(s) = K''_P + K''_D s$$

$$G_c(s) = (K''_P + K''_D s) \left( K'_P + \frac{K'_I}{s} \right)$$

# PID Control = PD + PI



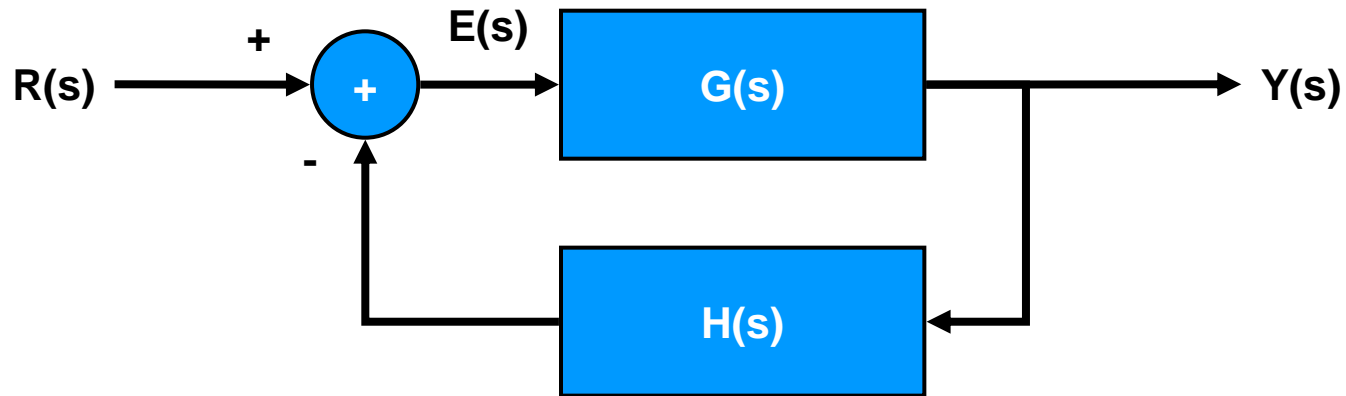
$$G_{PI}(s) = K'_P + \frac{K'_I}{s}$$

$$G_{PD}(s) = K''_P + K''_D s$$

$$G_c(s) = (K''_P + K''_D s) \left( K'_P + \frac{K'_I}{s} \right)$$

$$G_c(s) = (K'_P K''_P + K'_I K''_D) + \frac{K'_I K''_P}{s} + K'_P K''_D s$$

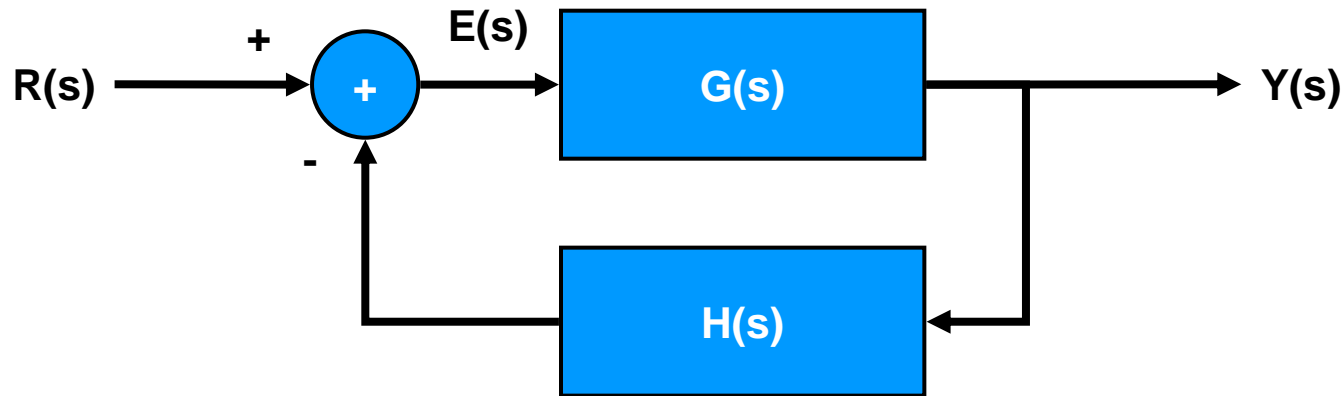
# Type 0, Type 1, Type 2,... Systems



$$G(s)H(s) = \frac{K(T_a s + 1)(T_b s + 1)\dots(T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1)\dots(T_p s + 1)}$$

**N is the number of integrations in the open loop transfer function**

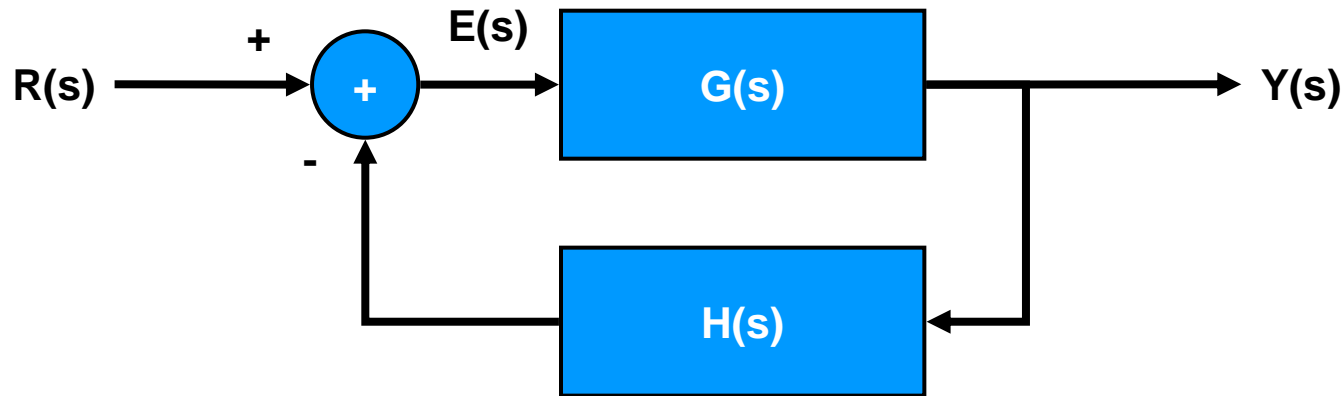
# Type 0, Type 1, Type 2,... Systems



$$G(s)H(s) = \frac{K(T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)}$$

$$\frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)} \quad \frac{E(s)}{R(s)} = 1 - \frac{Y(s)H(s)}{R(s)} = \frac{1}{1 + G(s)H(s)}$$

# Type 0, Type 1, Type 2,... Systems

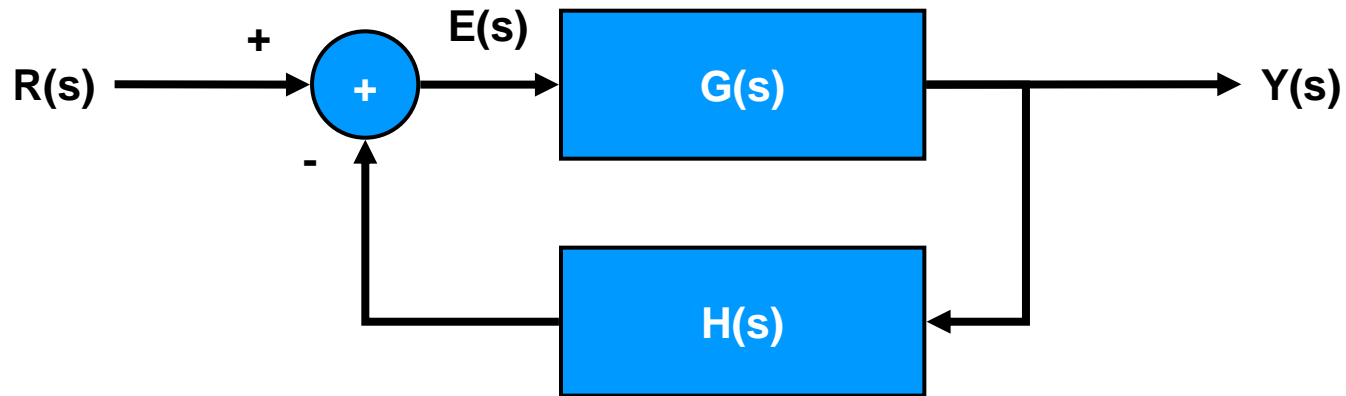


$$G(s)H(s) = \frac{K(T_a s + 1)(T_b s + 1)\dots(T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1)\dots(T_p s + 1)}$$

$$\frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)} \quad \frac{E(s)}{R(s)} = 1 - \frac{Y(s)H(s)}{R(s)} = \frac{1}{1 + G(s)H(s)}$$

$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

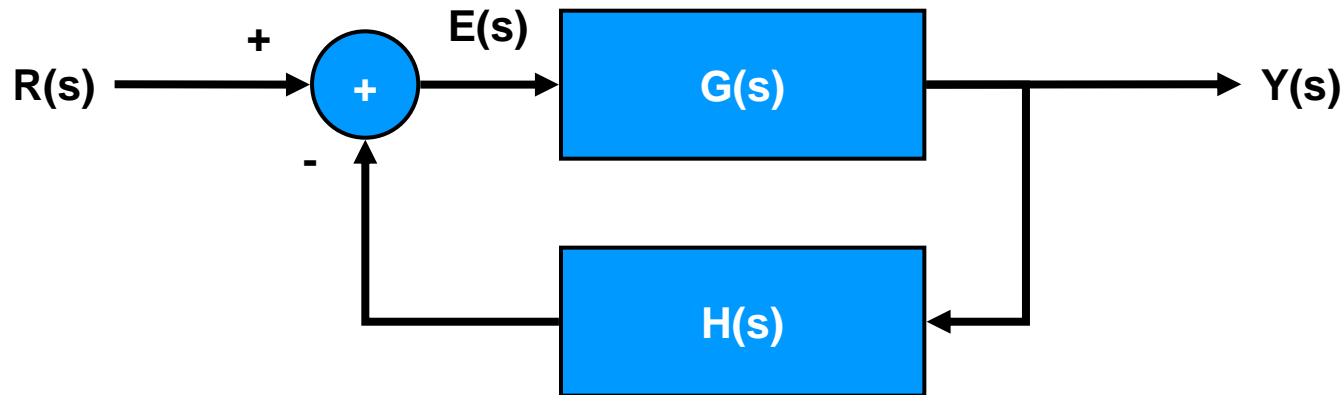
# Type 0, Type 1, Type 2,... Systems



$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

$$e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{sR(s)}{1 + G(s)H(s)}$$

# Type 0, Type 1, Type 2,... Systems



$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

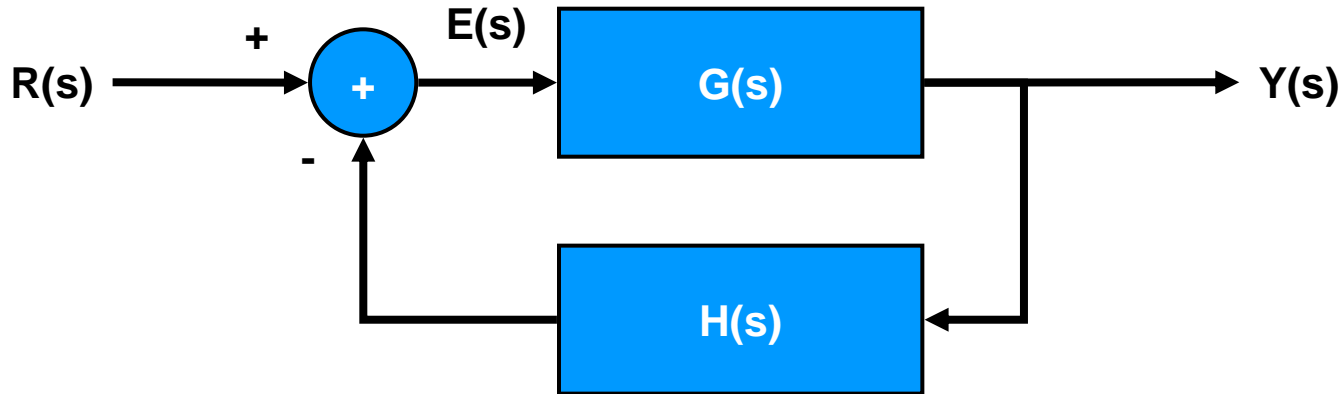
$$e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{sR(s)}{1 + G(s)H(s)}$$

**Static position error coefficient:**

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = G(0)H(0)$$

# Type 0, Type 1, Type 2,... Systems

- With unit step input  $r(t) = \begin{cases} 0 & t < 0 \\ 1 & t > 0 \end{cases}$



$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

$$e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{s}{1 + G(s)H(s)} \frac{1}{s}$$

**Static position error coefficient:**

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = G(0)H(0)$$

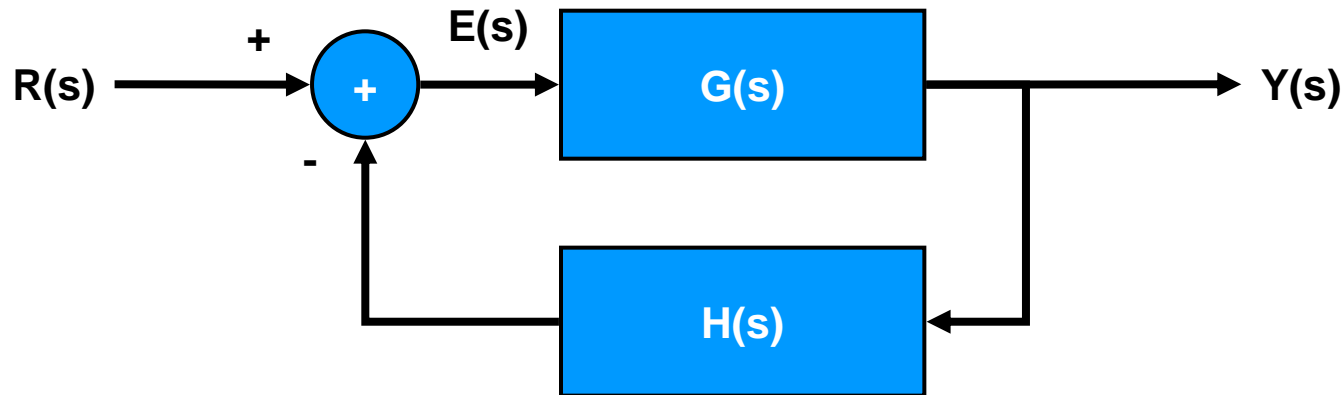
**For a unit step input:**

$$e_{ss} = \frac{1}{1 + K_p}$$

# Type 0, Type 1, Type 2,... Systems

- With unit step input

$$r(t) = \begin{cases} 0 & t < 0 \\ 1 & t > 0 \end{cases}$$



**Type 0:**

$$K_p = \lim_{s \rightarrow 0} \frac{K(T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{(T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} = K$$

$$e_{ss} = \frac{1}{1 + K} > 0$$

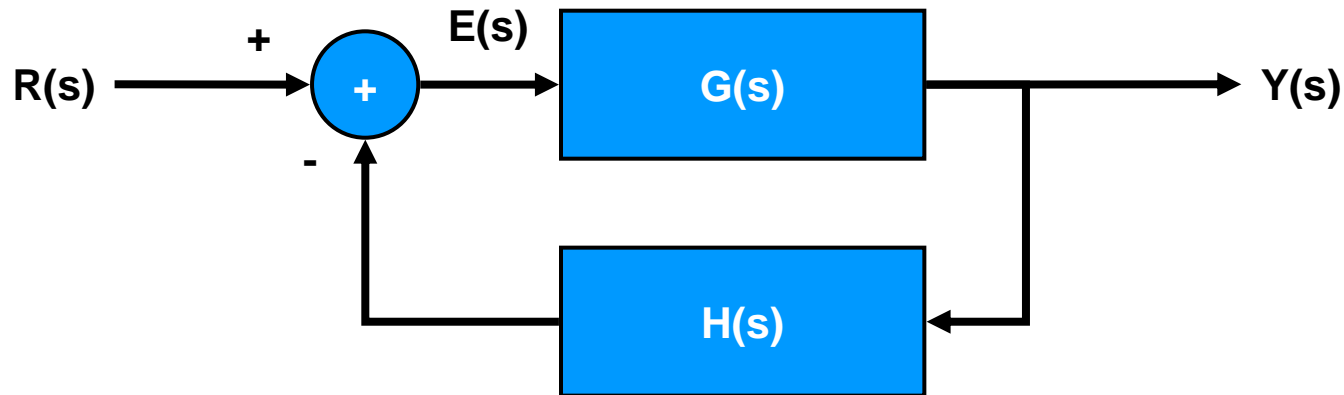
**Type 1 or higher:**

$$K_p = \lim_{s \rightarrow 0} \frac{K(T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} \rightarrow \infty$$

$$e_{ss} = 0$$

# Type 0, Type 1, Type 2,... Systems

- With unit ramp input  $r(t) = \begin{cases} 0 & t < 0 \\ t & t \geq 0 \end{cases}$



$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

$$e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{s}{1 + G(s)H(s)} \frac{1}{s^2}$$

**Static velocity error coefficient:**

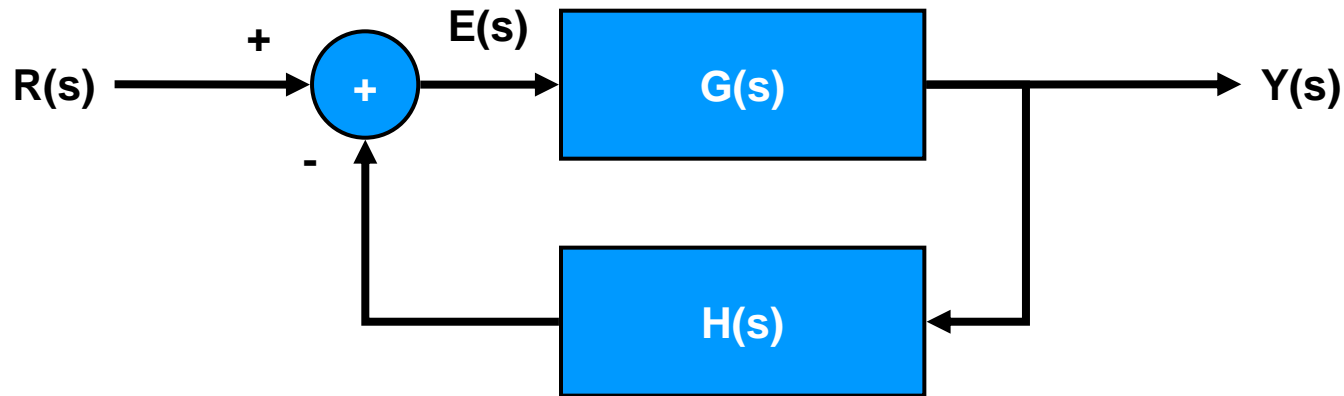
$$K_v = \lim_{s \rightarrow 0} sG(s)H(s)$$

**For a unit ramp input:**

$$e_{ss} = \frac{1}{K_v}$$

# Type 0, Type 1, Type 2,... Systems

- With unit ramp input  $r(t) = \begin{cases} 0 & t < 0 \\ t & t \geq 0 \end{cases}$



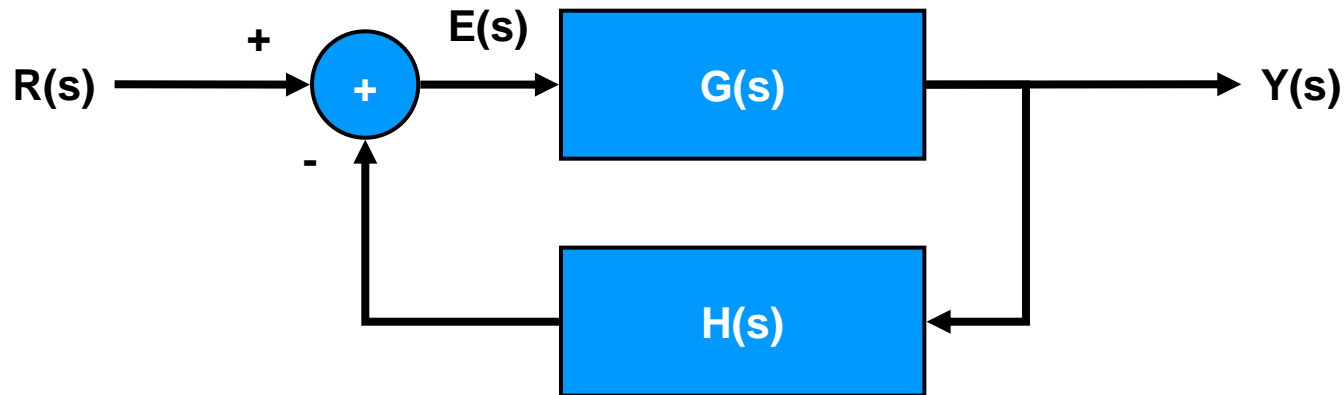
**Type 0:** 
$$K_v = \lim_{s \rightarrow 0} \frac{sK(T_a s + 1)(T_b s + 1)\dots(T_m s + 1)}{(T_1 s + 1)(T_2 s + 1)\dots(T_p s + 1)} = 0$$
 
$$e_{ss} = \frac{1}{K_v} \rightarrow \infty$$

**Type 1:** 
$$K_p = \lim_{s \rightarrow 0} \frac{sK(T_a s + 1)(T_b s + 1)\dots(T_m s + 1)}{s(T_1 s + 1)(T_2 s + 1)\dots(T_p s + 1)} = K$$
 
$$e_{ss} = \frac{1}{K_v} = \frac{1}{K}$$

**Type 2 or higher:** 
$$K_v = \lim_{s \rightarrow 0} \frac{sK(T_a s + 1)(T_b s + 1)\dots(T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1)\dots(T_p s + 1)} \rightarrow \infty$$
 
$$e_{ss} = \frac{1}{K_v} = 0$$

# Type 0, Type 1, Type 2,... Systems

- With unit parabolic input  $r(t) = \begin{cases} 0 & t < 0 \\ \frac{1}{2}t^2 & t \geq 0 \end{cases}$



$$E(s) = \frac{R(s)}{1 + G(s)H(s)}$$

$$e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{s}{1 + G(s)H(s)} \frac{1}{s^3}$$

**Static acceleration error coefficient:**

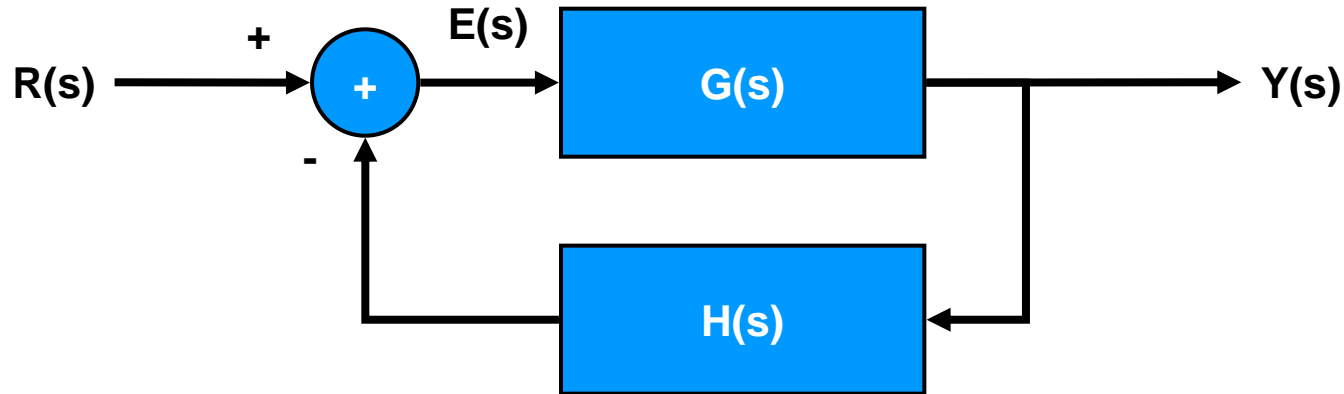
$$K_q = \lim_{s \rightarrow 0} s^2 G(s)H(s)$$

**For a unit parabolic input:**

$$e_{ss} = \frac{1}{K_a}$$

# Type 0, Type 1, Type 2,... Systems

- With unit parabolic input  $r(t) = \begin{cases} 0 & t < 0 \\ \frac{1}{2}t^2 & t \geq 0 \end{cases}$



**Type 0:** 
$$K_a = \lim_{s \rightarrow 0} \frac{s^2 K (T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{(T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} = 0$$
 
$$e_{ss} = \frac{1}{K_a} \rightarrow \infty$$

**Type 1:** 
$$K_a = \lim_{s \rightarrow 0} \frac{s^2 K (T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{s (T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} = 0$$
 
$$e_{ss} = \frac{1}{K_a} \rightarrow \infty$$

**Type 2:** 
$$K_a = \lim_{s \rightarrow 0} \frac{s^2 K (T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{s^2 (T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} = K$$
 
$$e_{ss} = \frac{1}{K}$$

**Type 3 or higher:** 
$$K_a = \lim_{s \rightarrow 0} \frac{s^2 K (T_a s + 1)(T_b s + 1) \dots (T_m s + 1)}{s^N (T_1 s + 1)(T_2 s + 1) \dots (T_p s + 1)} \rightarrow \infty$$
 
$$e_{ss} = \frac{1}{K_a} = 0$$

# Steady State Error

	Step Input $r(t) = 1$	Ramp input $r(t)=t$	Acceleration input $r(t) = \frac{1}{2} t^2$
Type 0 System	$\frac{1}{1+K}$	$\infty$	$\infty$
Type 1 System	0	$\frac{1}{K}$	$\infty$
Type 2 System	0	0	$\frac{1}{K}$

# Other Performance Indices

- First assume:

$$\lim_{t \rightarrow \infty} e(t) = 0$$

**Otherwise, other indices don't converge**

# Other Performance Indices

- First assume:

$$\lim_{t \rightarrow \infty} e(t) = 0$$

**Otherwise, other indices don't converge**

- Integral-square criterion

$$\int_0^{\infty} e^2(t) dt$$

- **Emphasizes large errors**
- **“Power” measurement**

# Other Performance Indices

- First assume:

$$\lim_{t \rightarrow \infty} e(t) = 0$$

**Otherwise, other indices don't converge**

- Integral-square criterion

$$\int_0^{\infty} e^2(t) dt$$

- **Emphasizes large errors**
- **“Power” measurement**

- Integral-of-time-multiplied square-error criterion

$$\int_0^{\infty} te^2(t) dt$$

- **Early errors are less important than later ones**

# Other Performance Indices

- First assume:

$$\lim_{t \rightarrow \infty} e(t) = 0$$

**Otherwise, other indices don't converge**

- Integral-square criterion

$$\int_0^{\infty} e^2(t) dt$$

- **Emphasizes large errors**
- **“Power” measurement**

- Integral-of-time-multiplied square-error criterion

$$\int_0^{\infty} t e^2(t) dt$$

- **Early errors are less important than later ones**

- Integral absolute-error criterion

$$\int_0^{\infty} |e(t)| dt$$

- **Not easy to compute analytically**
- **Good overall measure**

# Other Performance Indices

- First assume:

$$\lim_{t \rightarrow \infty} e(t) = 0$$

**Otherwise, other indices don't converge**

- Integral-square criterion

$$\int_0^{\infty} e^2(t) dt$$

- **Emphasizes large errors**
- **“Power” measurement**

- Integral-of-time-multiplied square-error criterion

$$\int_0^{\infty} t e^2(t) dt$$

- **Early errors are less important than later ones**

- Integral absolute-error criterion

$$\int_0^{\infty} |e(t)| dt$$

- **Not easy to compute analytically**
- **Good overall measure**

- Integral-of-time-multiplied absolute-error criterion

$$\int_0^{\infty} t |e(t)| dt$$

- **Not easy to compute analytically**
- **Early errors are less important than later ones**