

# Design IV

## E232 Fall 07

### Class 11

Bruce McNair  
bmcnair@stevens.edu

# Spectral Analysis With Arbitrary Signals

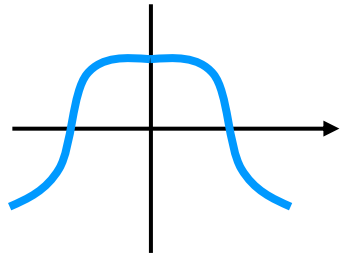
- Any well-behaved periodic signal  $f(t)$  can be represented as

$$f(t) = a_0 + \left( \sum_{i=1}^{\infty} a_i \cos\left(i \frac{2\pi}{T} t\right) \right) + \left( \sum_{i=1}^{\infty} b_i \sin\left(i \frac{2\pi}{T} t\right) \right)$$

where

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

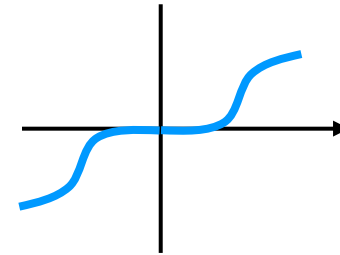
**Even function**



$$a_k = \frac{2}{T} \int_0^T f(t) \cos\left(k \frac{2\pi}{T} t\right) dt$$

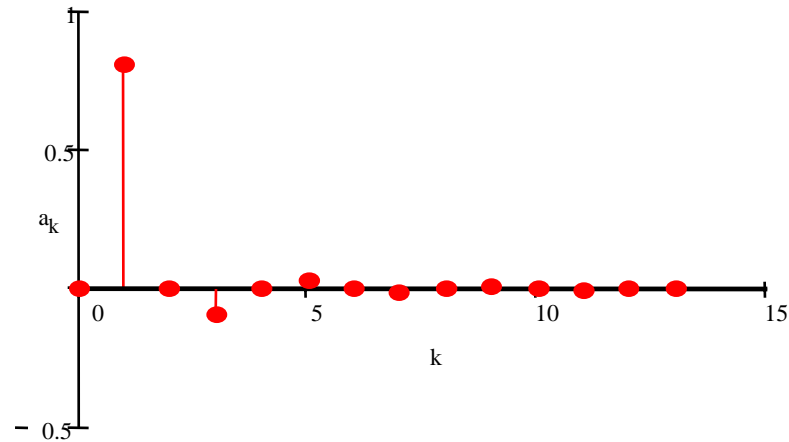
$$b_k = \frac{2}{T} \int_0^T f(t) \sin\left(k \frac{2\pi}{T} t\right) dt$$

**Odd function**

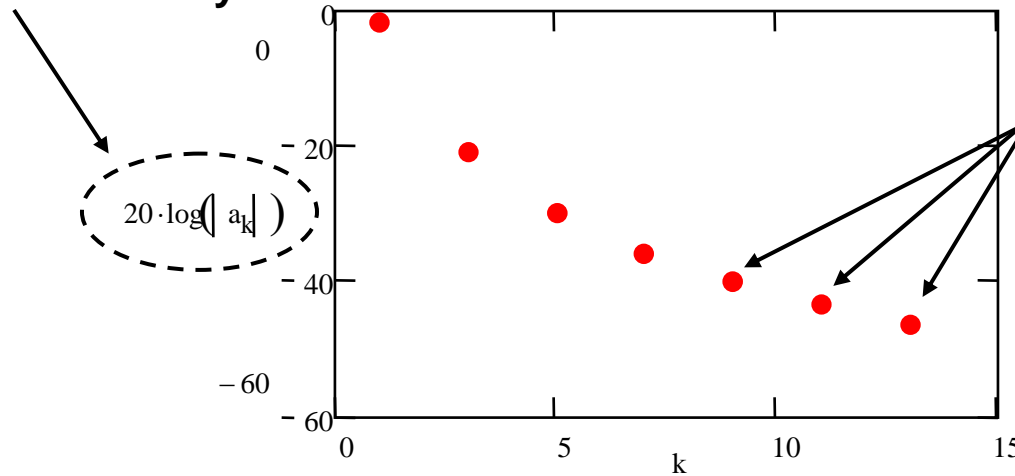


# Spectral Analysis With Arbitrary Signals

- Spectrum of triangular wave



Sine components only



Odd harmonics only

# Generalizing Fourier Series

$$f(t) = a_0 + \left( \sum_{i=1}^{\infty} a_i \cos\left(i \frac{2\pi}{T} t\right) \right) + \left( \sum_{i=1}^{\infty} b_i \sin\left(i \frac{2\pi}{T} t\right) \right)$$

**Fourier  
series**

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

$$a_k = \frac{2}{T} \int_0^T f(t) \cos\left(i \frac{2\pi}{T} t\right) dt$$

$$b_k = \frac{2}{T} \int_0^T f(t) \sin\left(i \frac{2\pi}{T} t\right) dt$$

# Generalizing Fourier Series

$$f(t) = a_0 + \left( \sum_{i=1}^{\infty} a_i \cos\left(i \frac{2\pi}{T} t\right) \right) + \left( \sum_{i=1}^{\infty} b_i \sin\left(i \frac{2\pi}{T} t\right) \right)$$

**Fourier  
series**

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

$$a_k = \frac{2}{T} \int_0^T f(t) \cos\left(i \frac{2\pi}{T} t\right) dt$$

$$b_k = \frac{2}{T} \int_0^T f(t) \sin\left(i \frac{2\pi}{T} t\right) dt$$

**Euler's  
formula**

$$e^{j\theta} = \cos(\theta) + j \sin(\theta)$$

$$\sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

$$\cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2}$$

# Generalizing Fourier Series

$$f(t) = a_0 + \left( \sum_{i=1}^{\infty} a_i \cos\left(i \frac{2\pi}{T} t\right) \right) + \left( \sum_{i=1}^{\infty} b_i \sin\left(i \frac{2\pi}{T} t\right) \right)$$

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j \frac{2\pi n t}{T}}$$

**Fourier series**

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

$$a_k = \frac{2}{T} \int_0^T f(t) \cos\left(i \frac{2\pi}{T} t\right) dt$$

$$b_k = \frac{2}{T} \int_0^T f(t) \sin\left(i \frac{2\pi}{T} t\right) dt$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j \frac{2\pi k t}{T}} dt$$

$$e^{j\theta} = \cos(\theta) + j \sin(\theta)$$

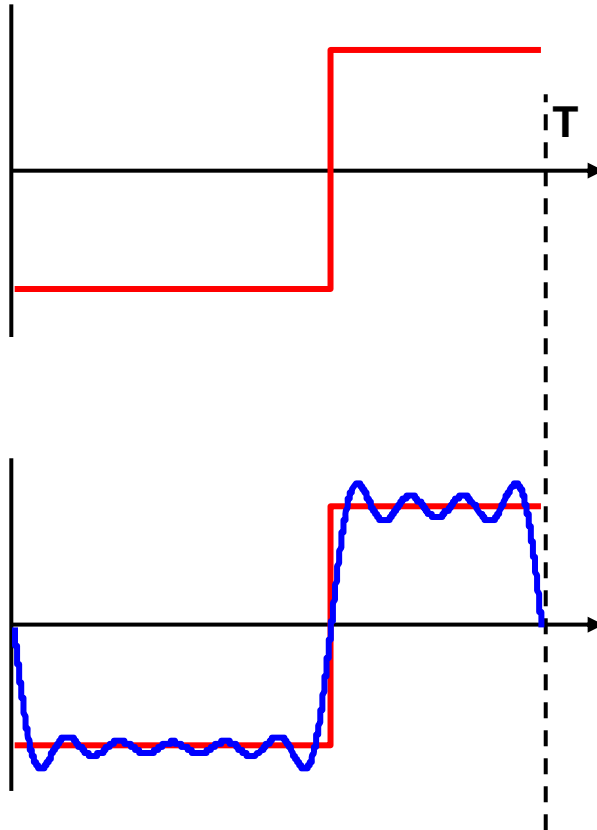
**Euler's formula**

$$\sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

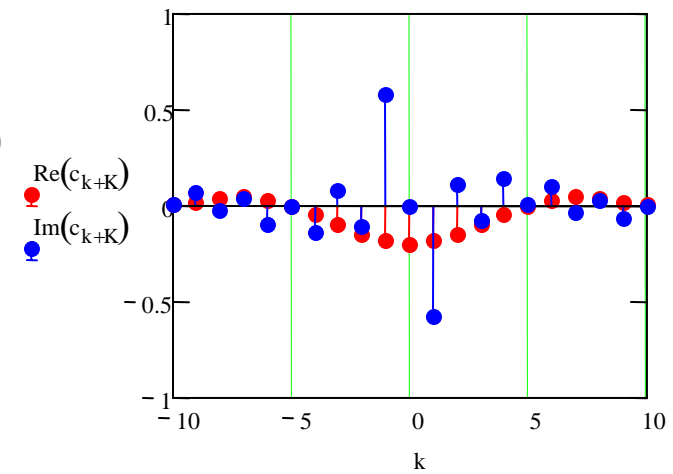
$$\cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2}$$

# Complex Spectrum Of A Signal

- Shifted square wave



**$N = 19$**



# Today's topics

- More on Frequency Domain analysis
- Introduction to control systems

# Generalizing The Fourier Series

- Start with the complex Fourier Series

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

# Generalizing The Fourier Series

- Change variables

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

**Replace  $2\pi/T$  with  $\omega_0$**

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jk\omega_0 t} dt$$

# Generalizing The Fourier Series: The Fourier Transform

- Consider what happens when the analysis period is allowed to increase

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

Replace  $2\pi/T$  with  $\omega_0$

Let  $\omega_0$  go to 0  
T becomes infinite

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jk\omega_0 t} dt$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

# Applying The Fourier Transform To Sampled Signals

- Continuous samples of  $f(t)$ ,  $F(\omega)$  are not available

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

# Applying The Fourier Transform To Sampled Signals

- Continuous samples of  $f(t)$ ,  $F(\omega)$  are not available

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$f(n\Delta t) = \frac{1}{N} \sum_{k=0}^{N-1} F(k\Delta f) e^{j(2\pi k\Delta f)(n\Delta t)}$$

Discrete time

Discrete frequency

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$F(k\Delta f) = \sum_{n=0}^{N-1} f(n\Delta t) e^{-j(2\pi k\Delta f)(n\Delta t)}$$

# Applying The Fourier Transform To Sampled Signals

- Continuous samples of  $f(t)$ ,  $F(\omega)$  are not available

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$f(n\Delta t) = \frac{1}{N} \sum_{k=0}^{N-1} F(k\Delta f) e^{j(2\pi k\Delta f)(n\Delta t)}$$

Discrete time

Discrete frequency

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$F(k\Delta f) = \sum_{n=0}^{N-1} f(n\Delta t) e^{-j(2\pi k\Delta f)(n\Delta t)}$$

## N-point Discrete Fourier Transform (DFT)

# Applying The Fourier Transform To Sampled Signals

- Continuous samples of  $f(t)$ ,  $F(\omega)$  are not available

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$f(n\Delta t) = \frac{1}{N} \sum_{k=0}^{N-1} F(k\Delta f) e^{j(2\pi k\Delta f)(n\Delta t)}$$

Discrete time

Discrete frequency

Note symmetry of  $e^{jx}$

Not all  $N^2$  factors need be calculated

$$F(k\Delta f) = \sum_{n=0}^{N-1} f(n\Delta t) e^{-j(2\pi k\Delta f)(n\Delta t)}$$

~~N-point Discrete Fourier Transform (DFT)~~

If  $N=2^M$ , ( $N \times \log(N)$ ) operations needed for Fast Fourier Transform (FFT)

# Examples of Control Systems

Hard disk read/write head positioning

CNC machining tools

Water treatment system

Car's cruise control

Aircraft auto-pilot

Body's glucose level

Highway traffic control

Employee performance appraisal process

Car's engine controller

Modem's adaptive equalizer

Building heating/cooling system control

Corporate budgets

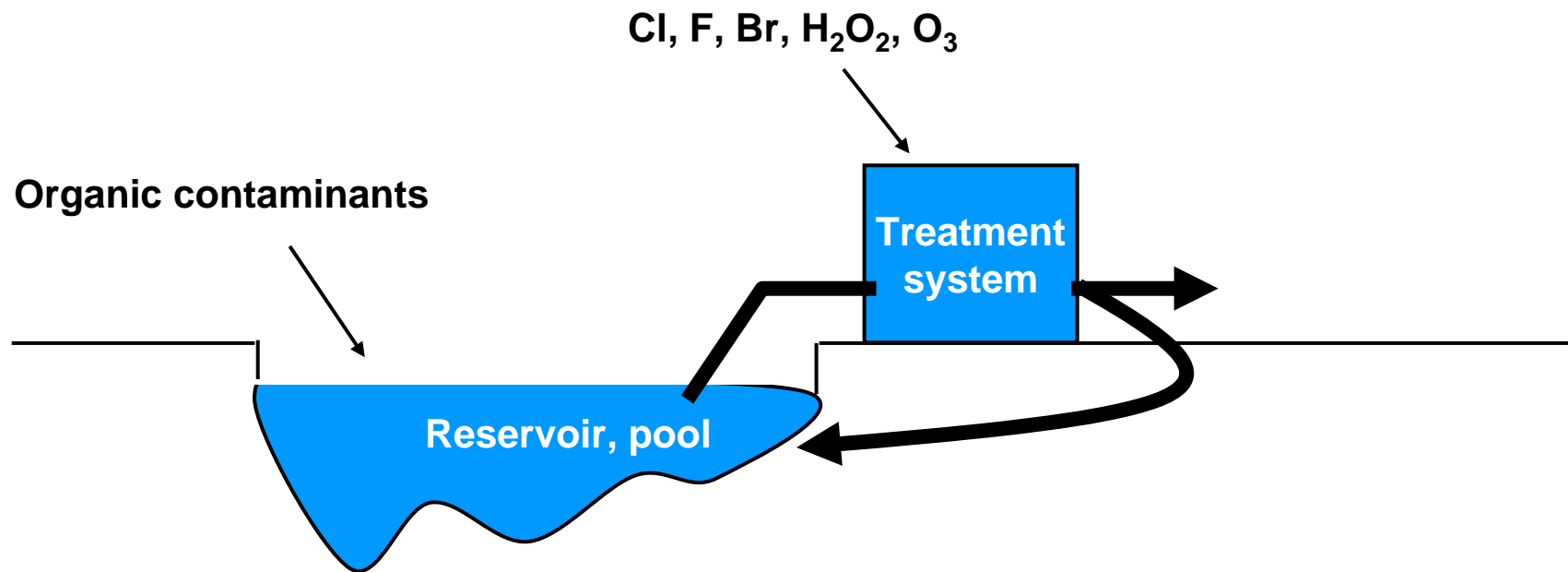
Body's heart-rate pacemaker

# Reference Material

- R. Dorf, R. Bishop, “Modern Control Engineering, 10<sup>th</sup> Ed.,” Pearson/Prentice-Hall, 2005, ISBN 0-13-145733-0.

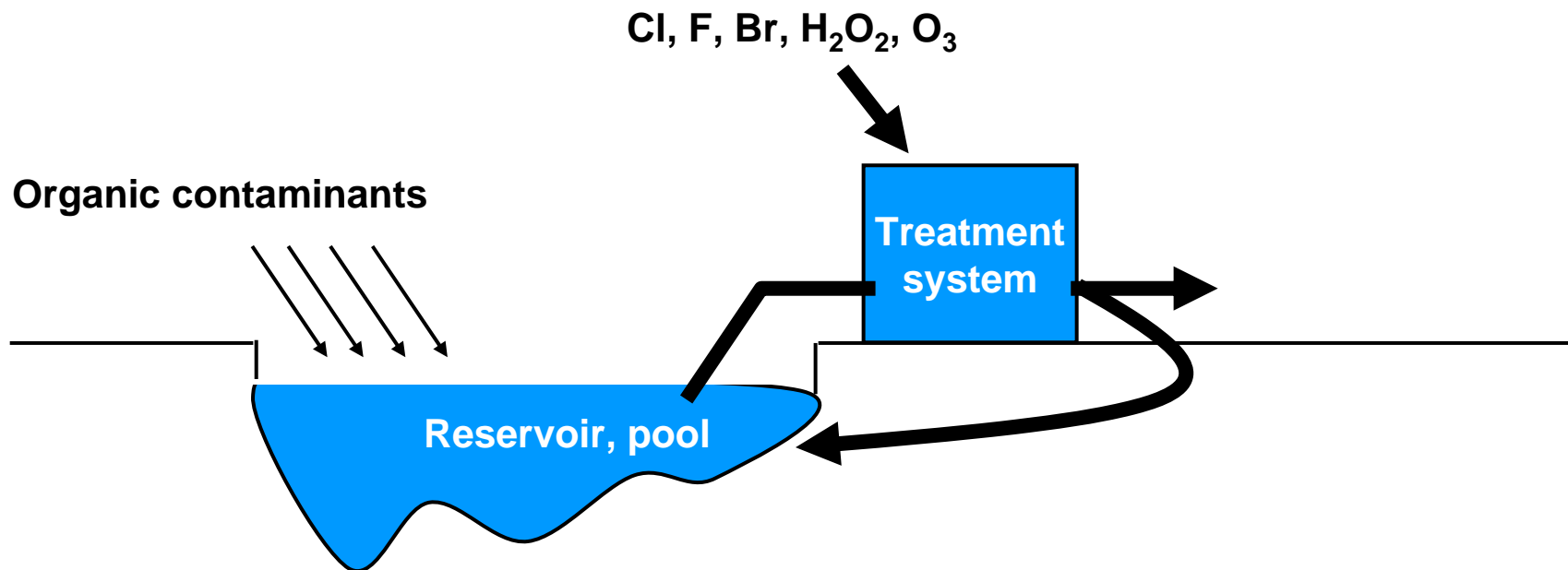
# Consider a Water Treatment System

- Chlorine, etc., are added to water to kill bacteria, deal with other organic contaminants



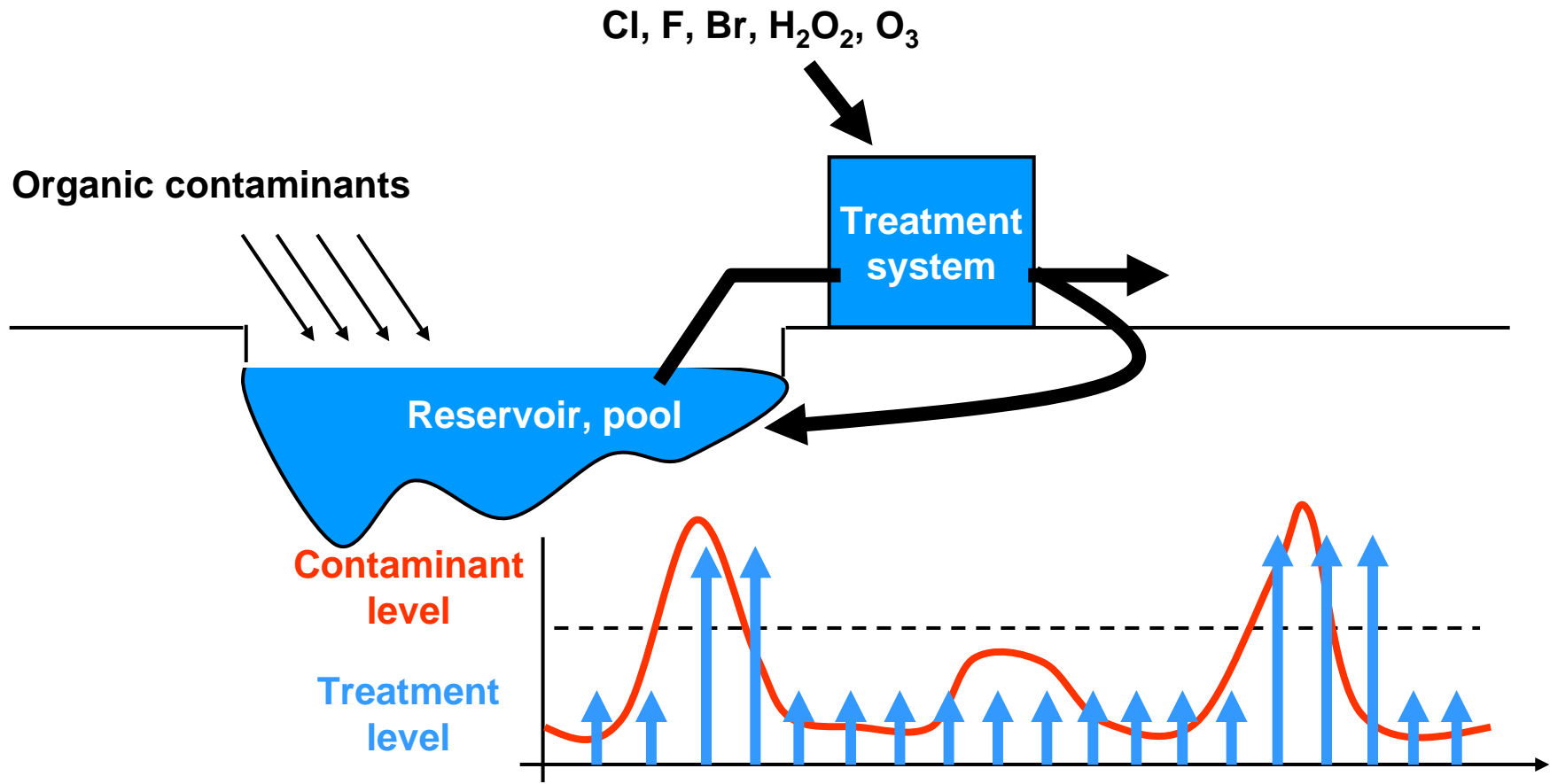
# Consider a Water Treatment System

- Chlorine, etc., are added to water to kill bacteria, deal with other organic contaminants
- Rain, runoff, usage may add additional contaminants, so we increase level of treatment



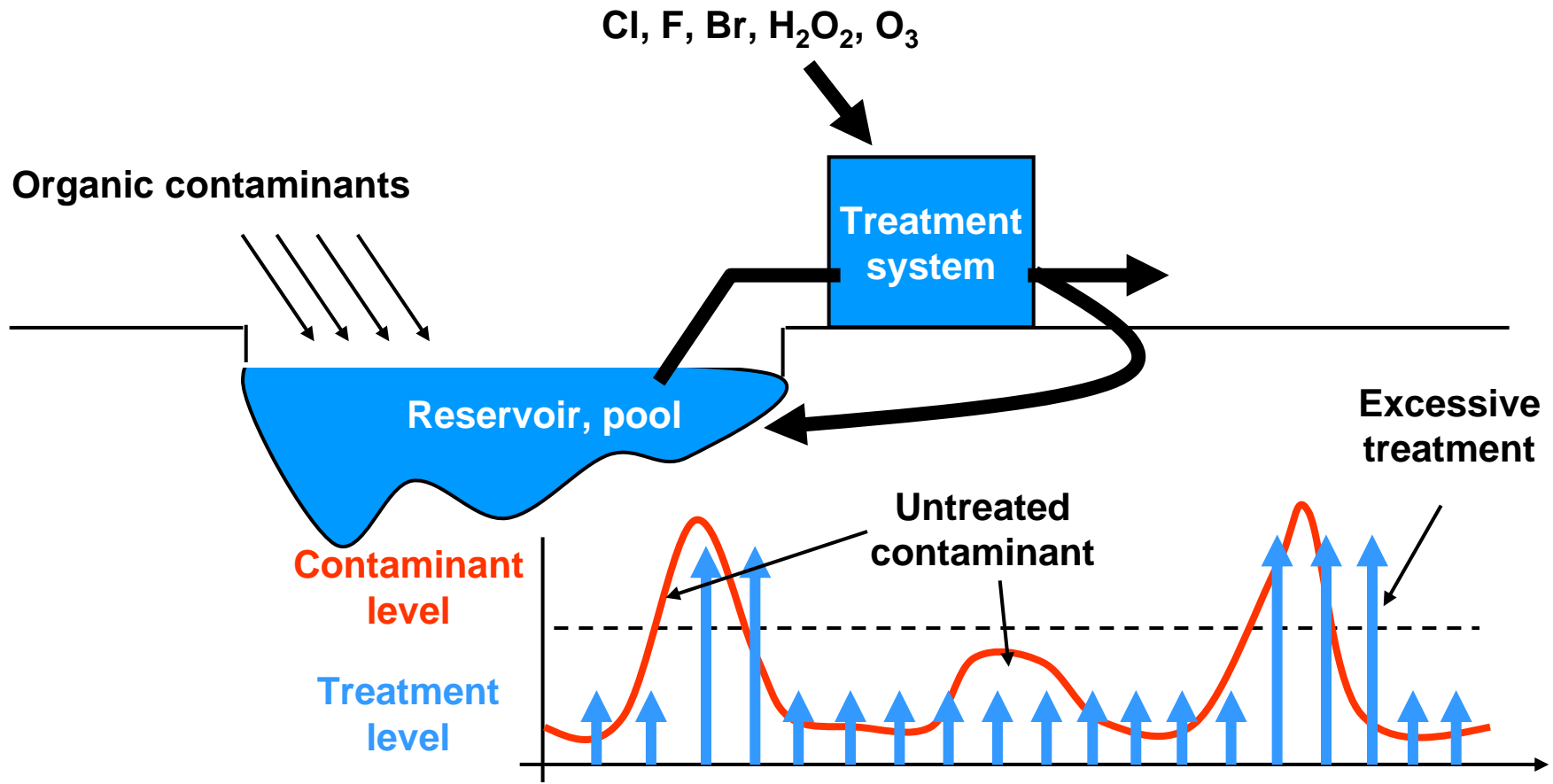
# Consider a Water Treatment System

- Chlorine, etc., are added to water to kill bacteria, deal with other organic contaminants
- Rain, runoff, usage may add additional contaminants, so we increase level of treatment



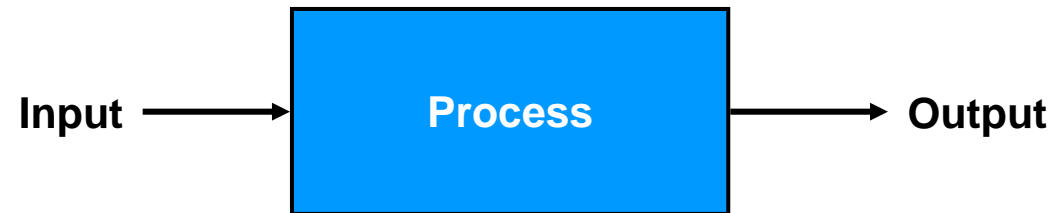
# Consider a Water Treatment System

- Chlorine, etc., are added to water to kill bacteria, deal with other organic contaminants
- Rain, runoff, usage may add additional contaminants, so we increase level of treatment



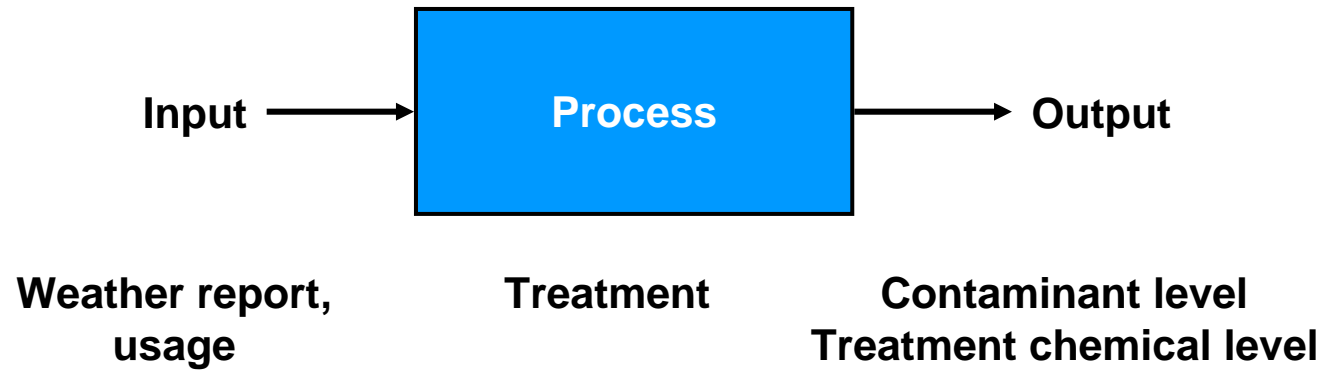
# Process Control

- Generic process control



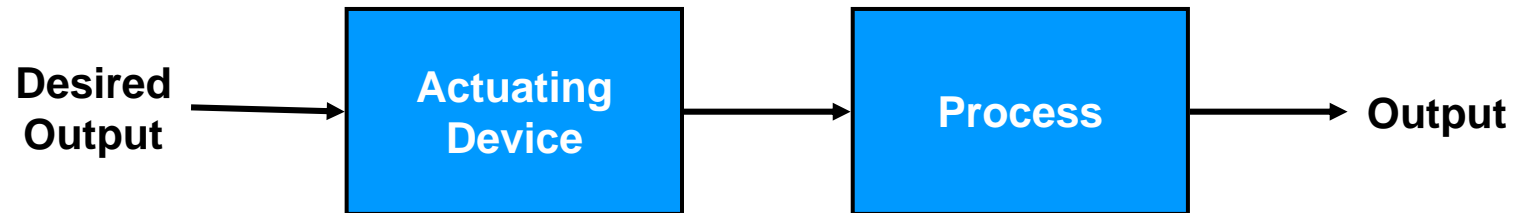
# Process Control

- Water treatment



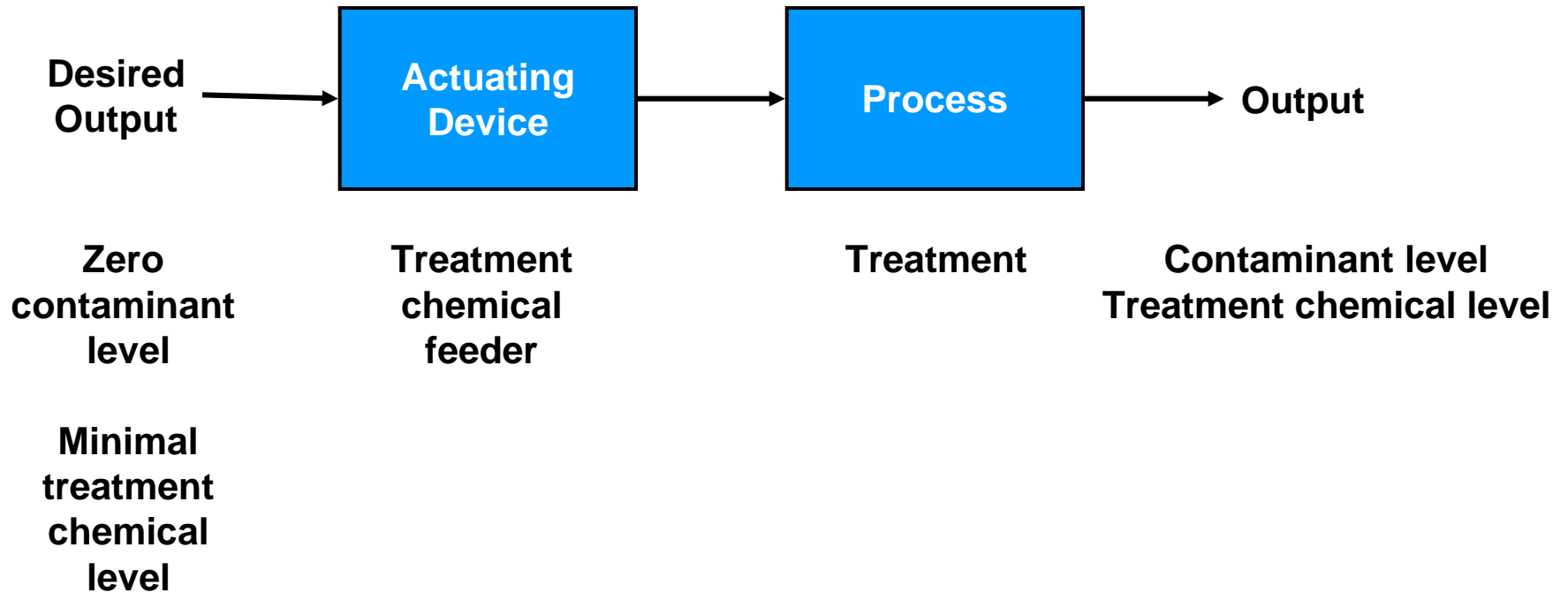
# Open Loop Process Control

- Generic process



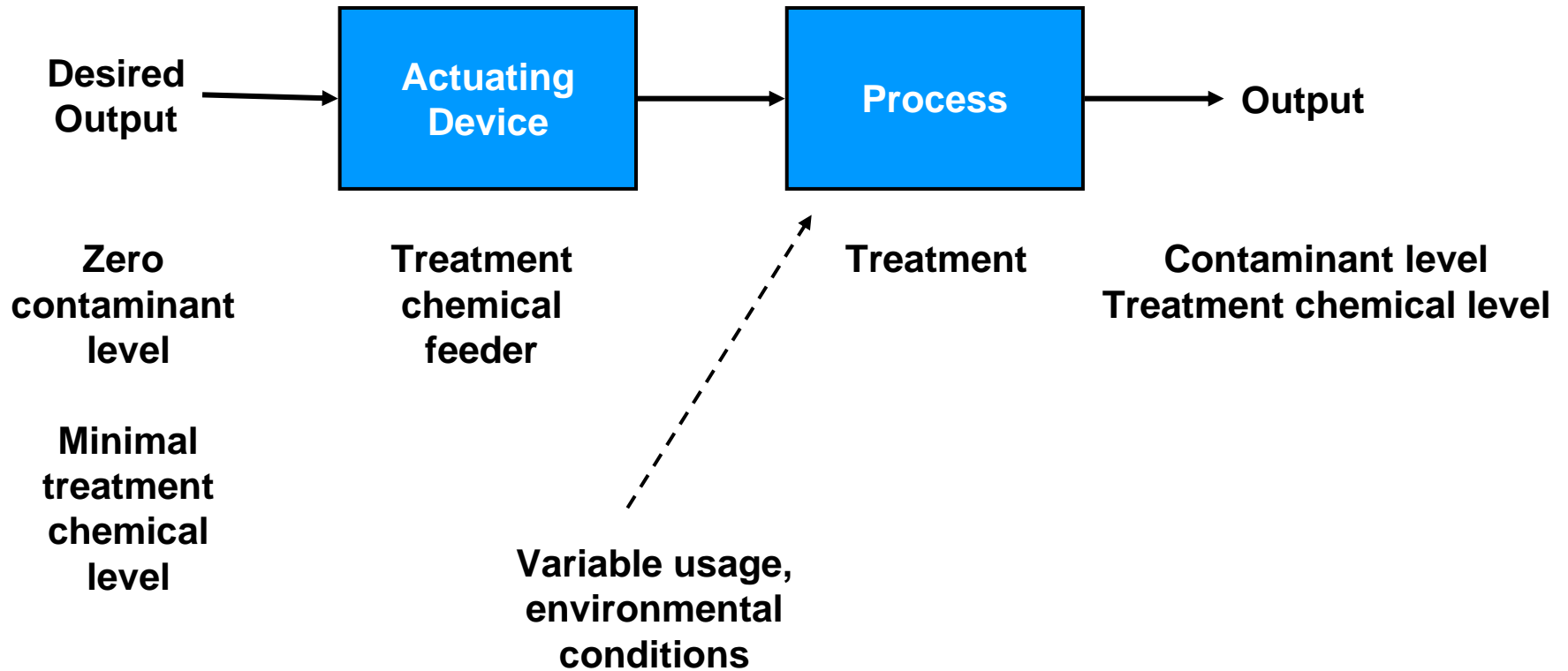
# Open Loop Process Control

- Water treatment



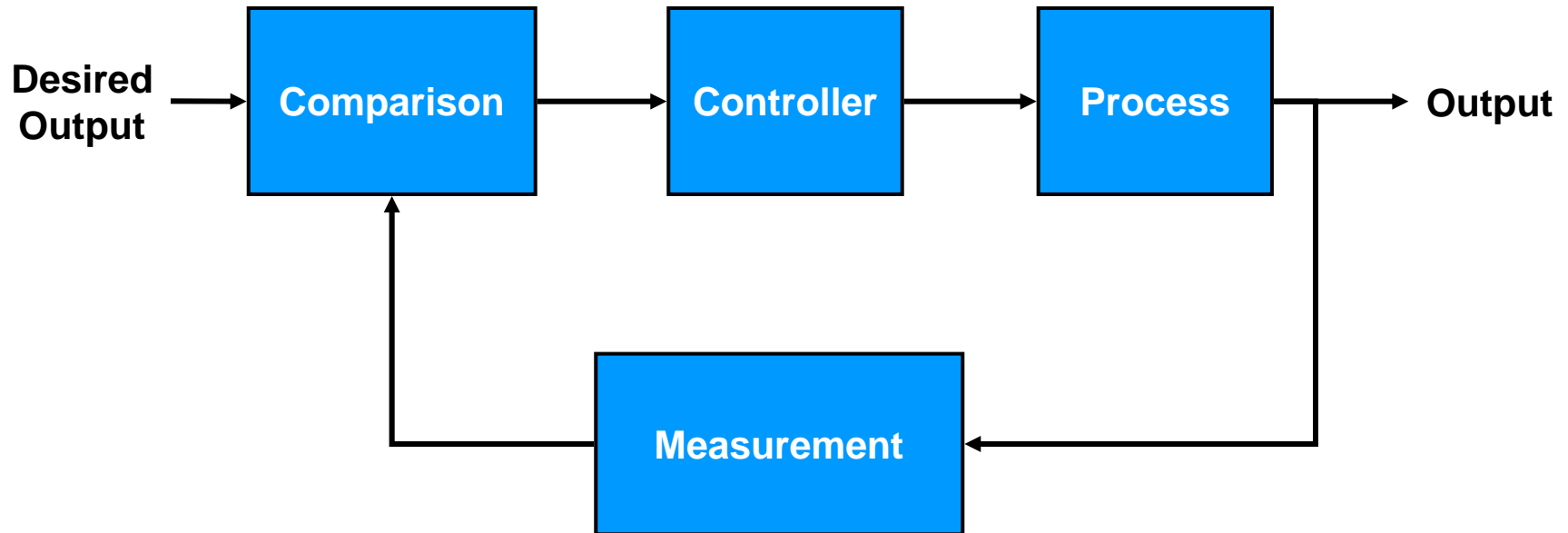
# Open Loop Process Control

- Water treatment



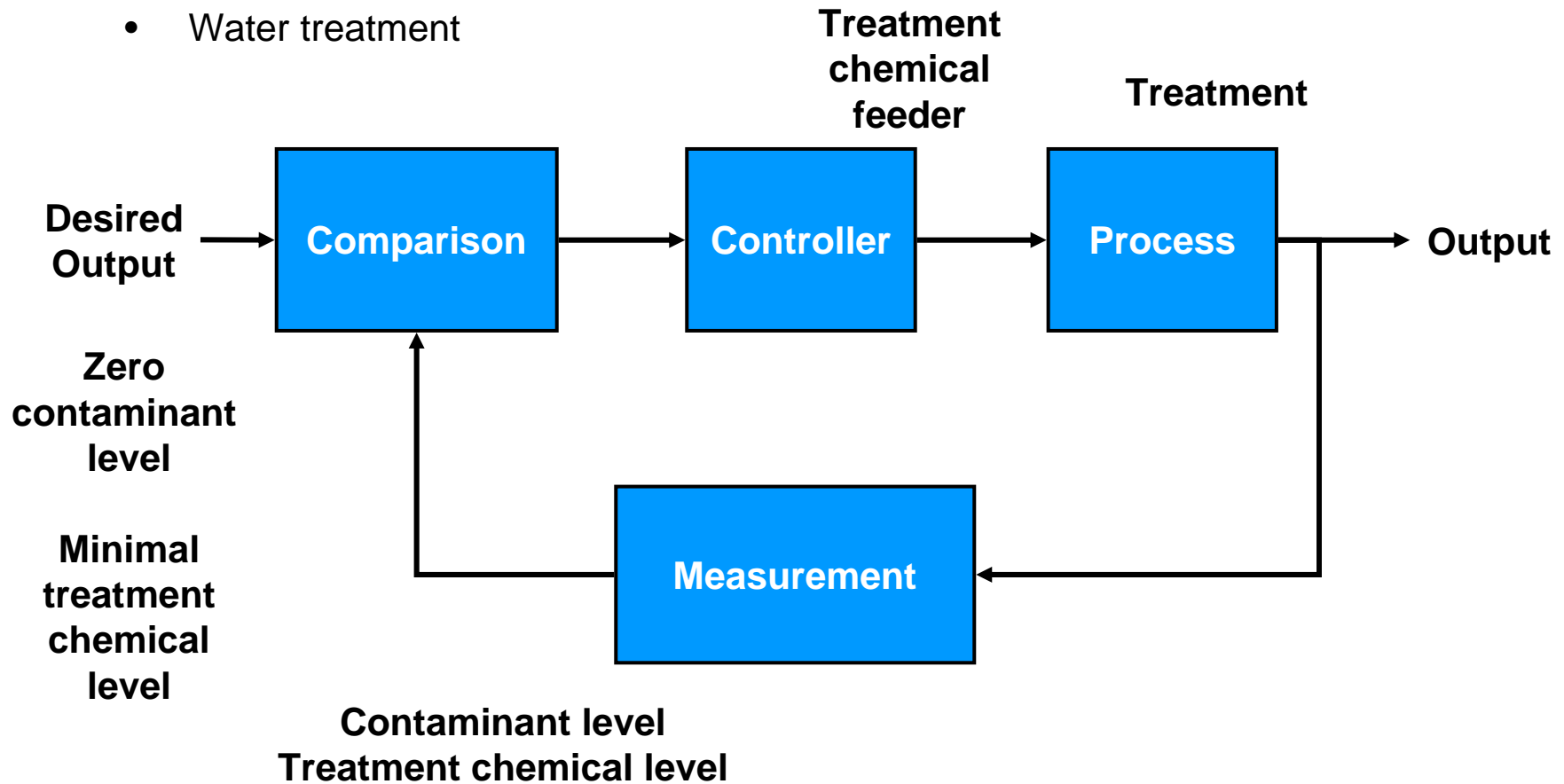
# Closed Loop Process Control

- Generic process



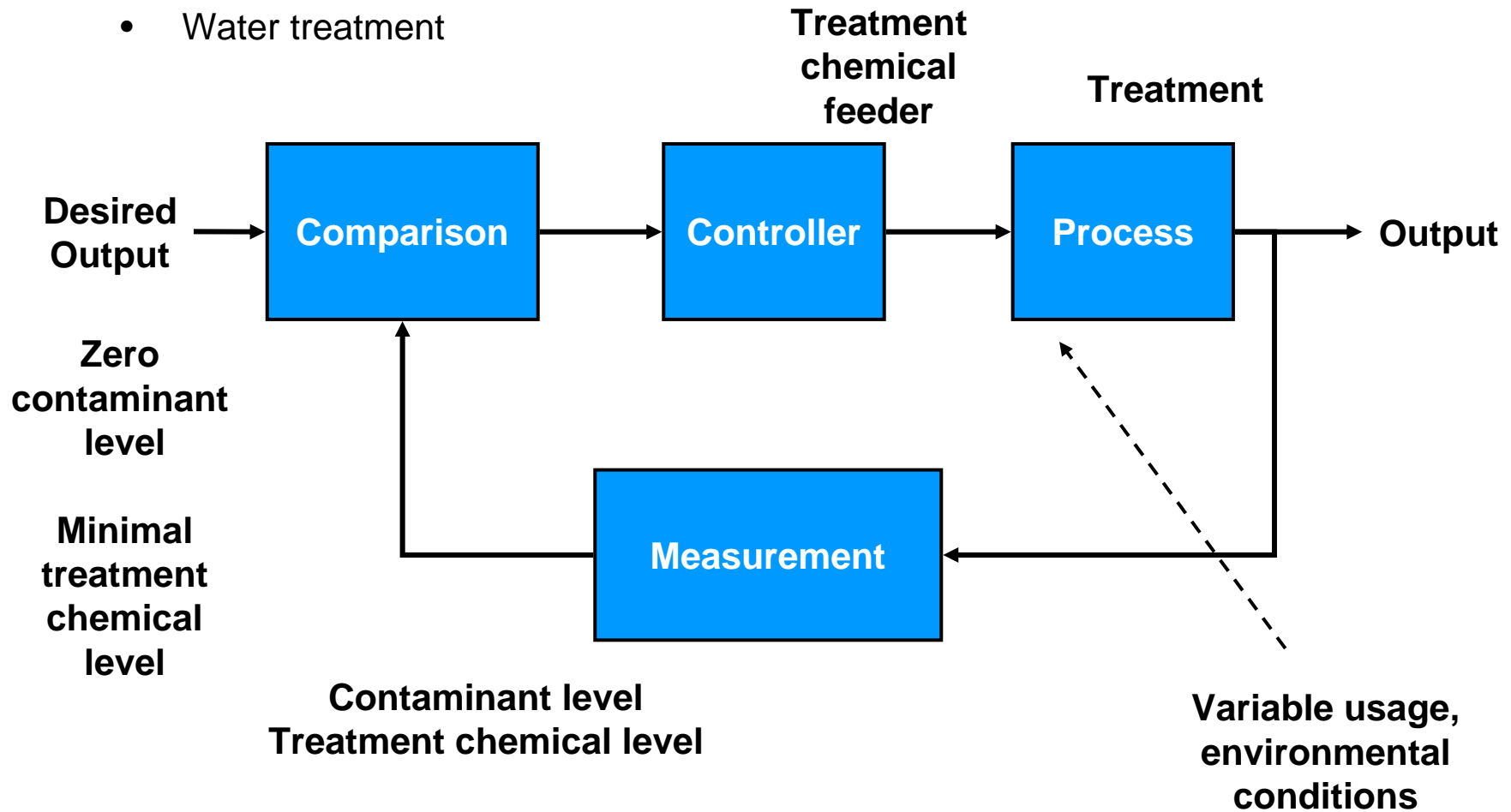
# Closed Loop Process Control

- Water treatment



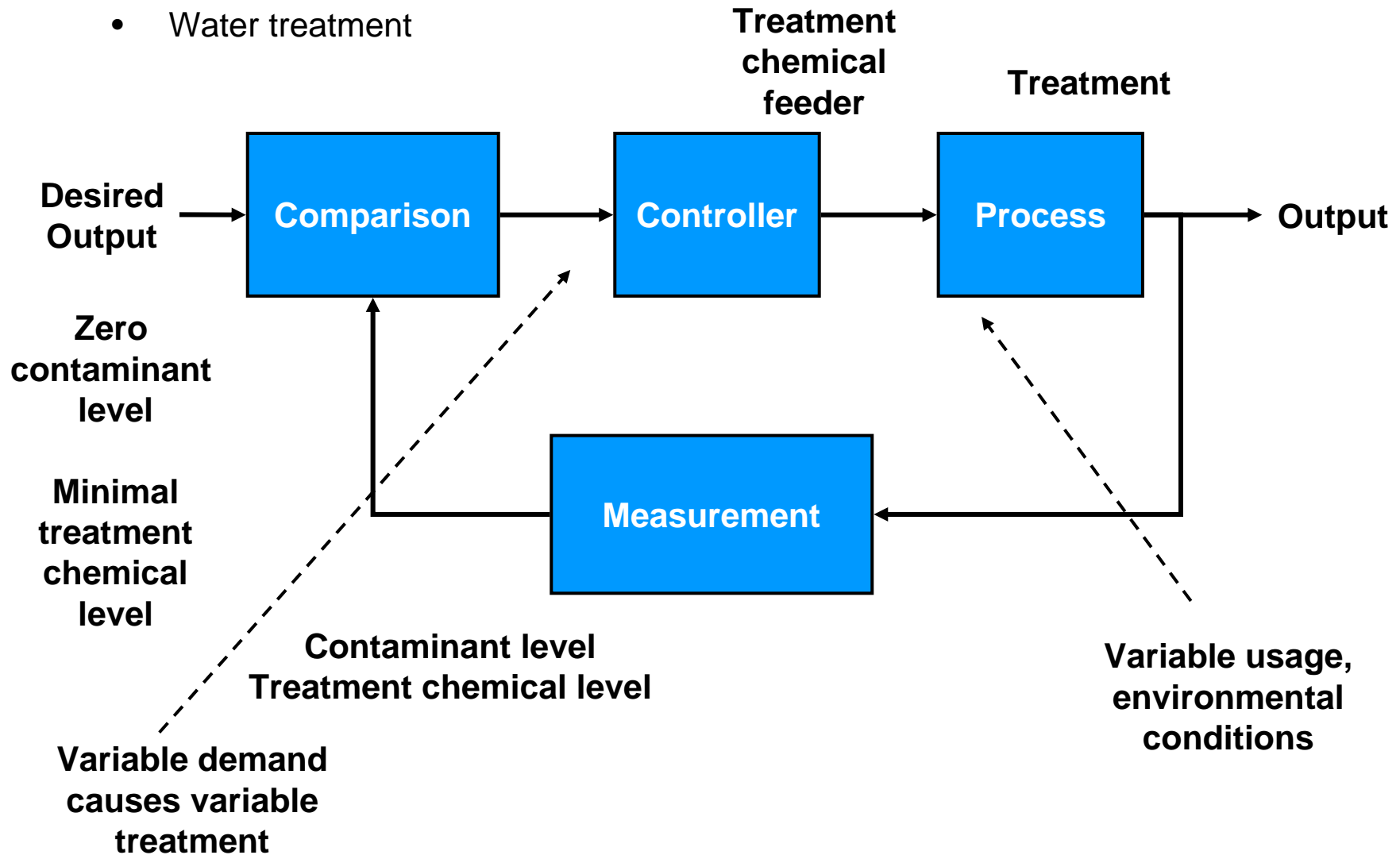
# Closed Loop Process Control

- Water treatment



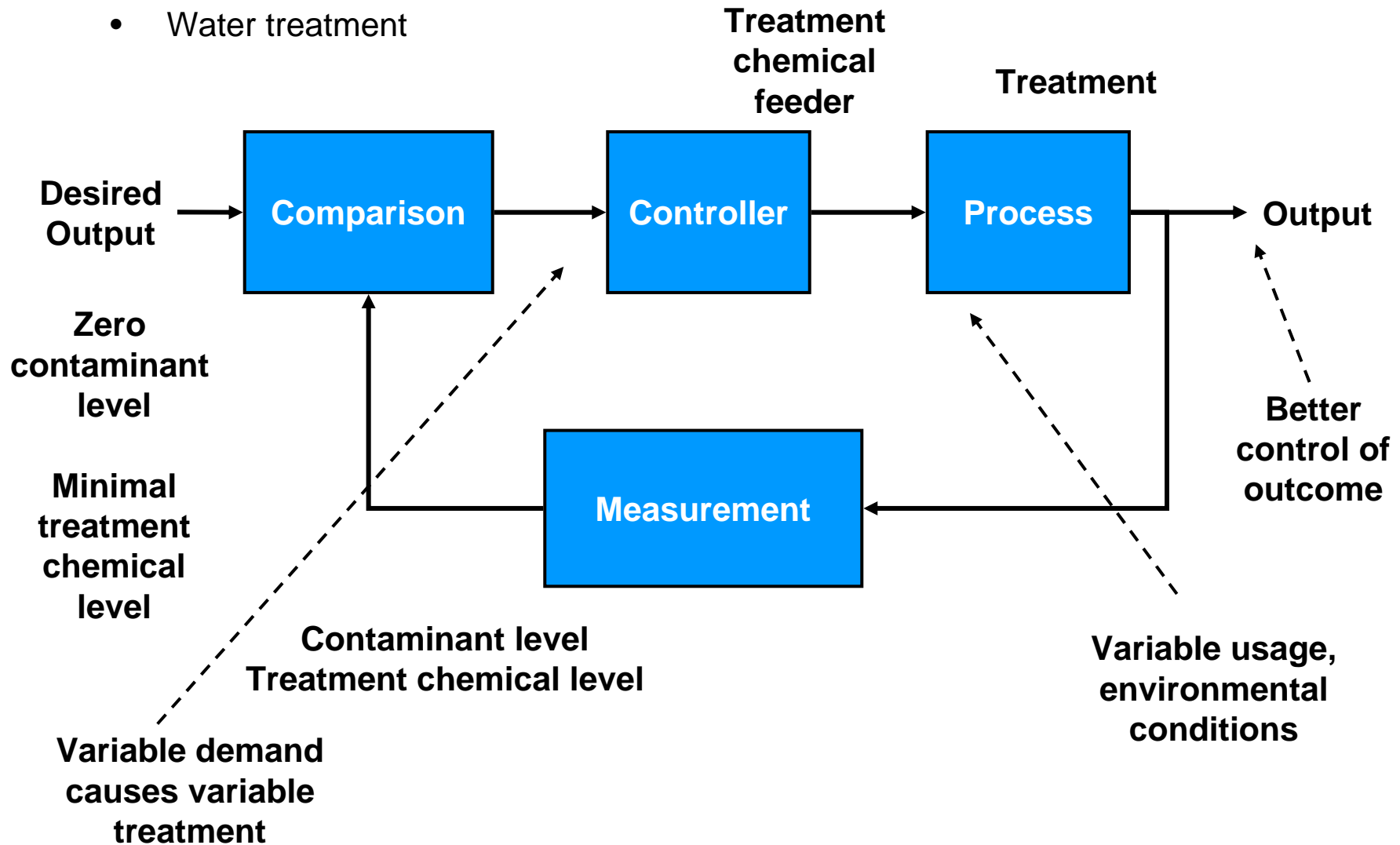
# Closed Loop Process Control

- Water treatment



# Closed Loop Process Control

- Water treatment



# History of Automatic Control

- 250 BC: Greek's use float regulator to control level of oil in lamps
- 1600's – 1700's: temperature, pressure regulators for steam boilers
- 1769: Watt's flyball speed governor for steam engine
- 1868: Maxwell's mathematical analysis of steam engine governor

Early 1900s: Telephone network

Mid 1900s: gun control

US: Frequency domain techniques (Laplace Transform)

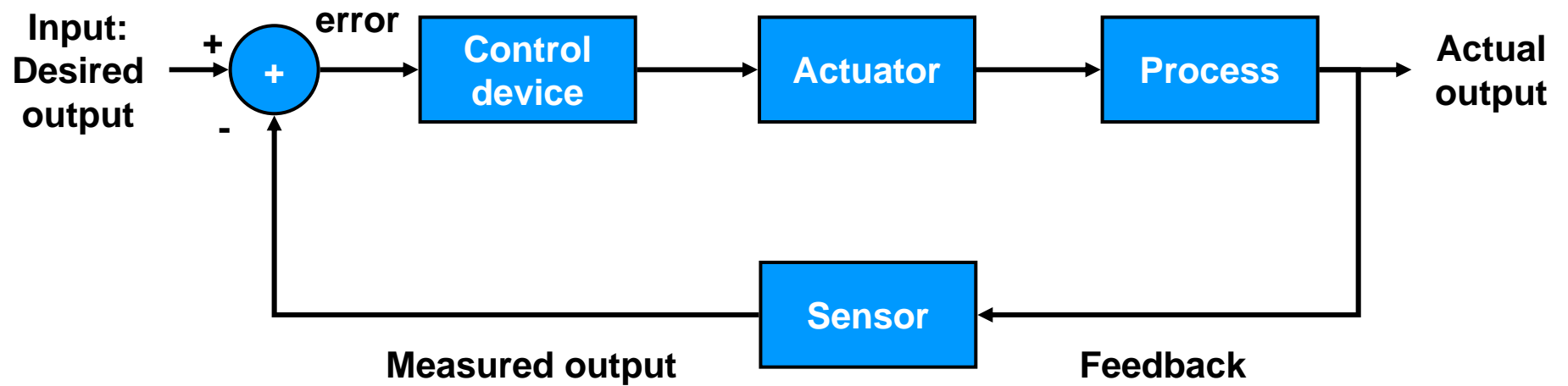
Mid 1900s: flight control, radar

Russia: Time domain techniques (differential equations)

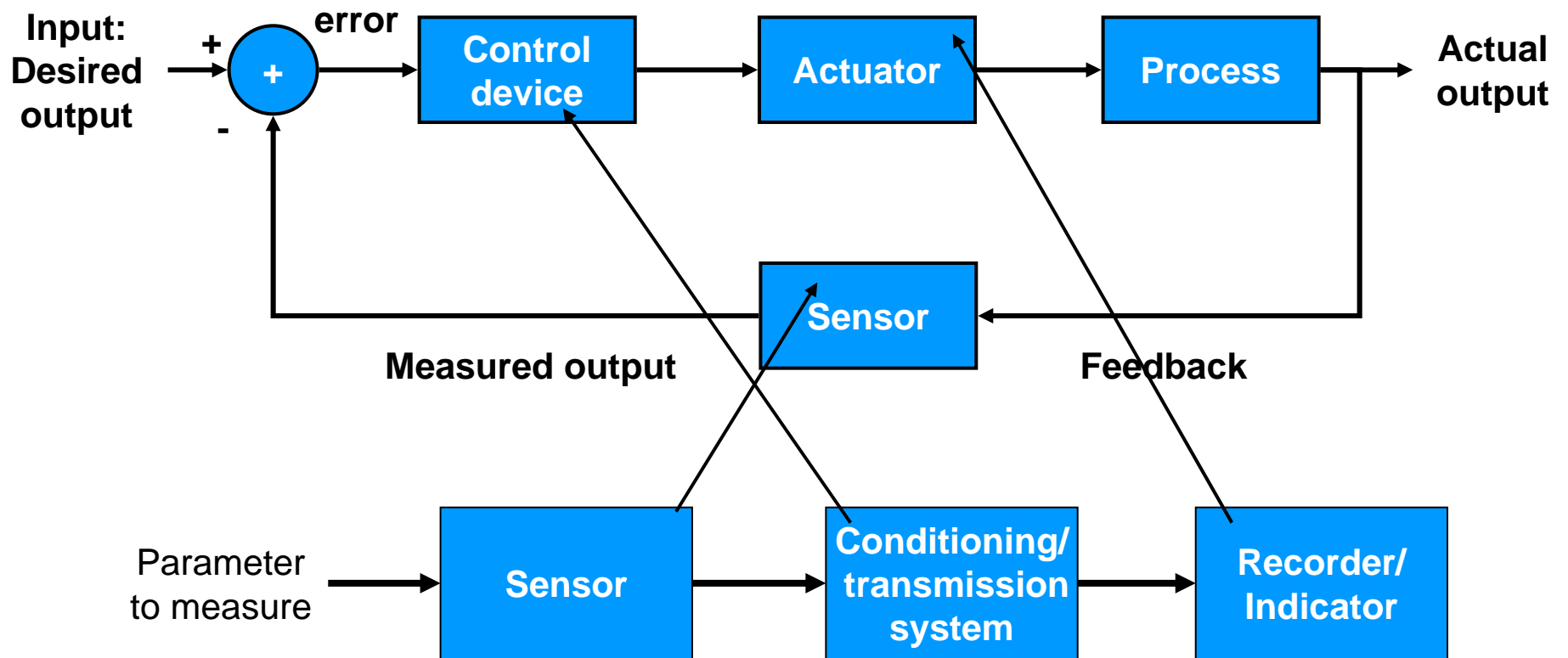
*Empirical design*

*Analytical design*

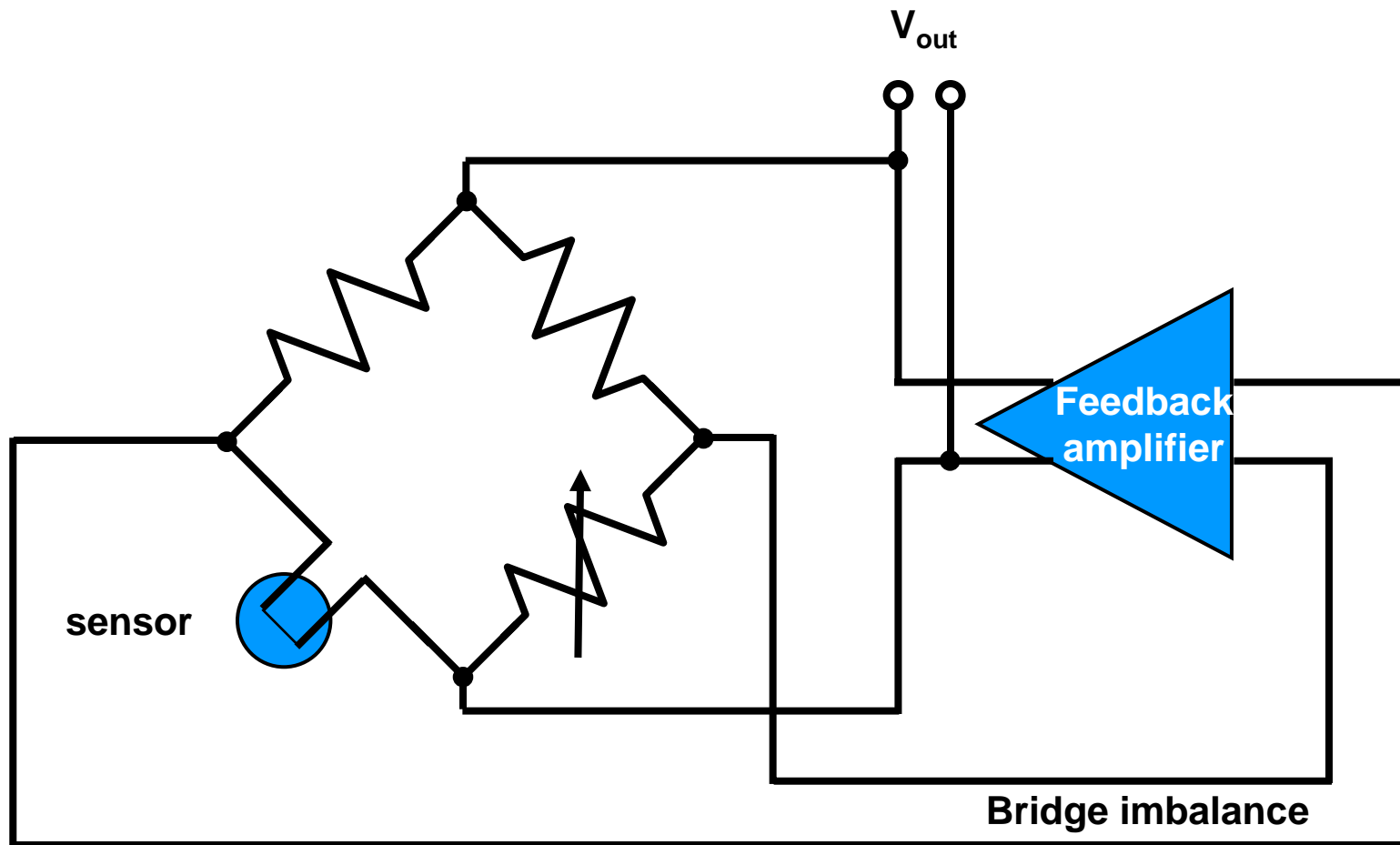
# Negative Feedback Control System



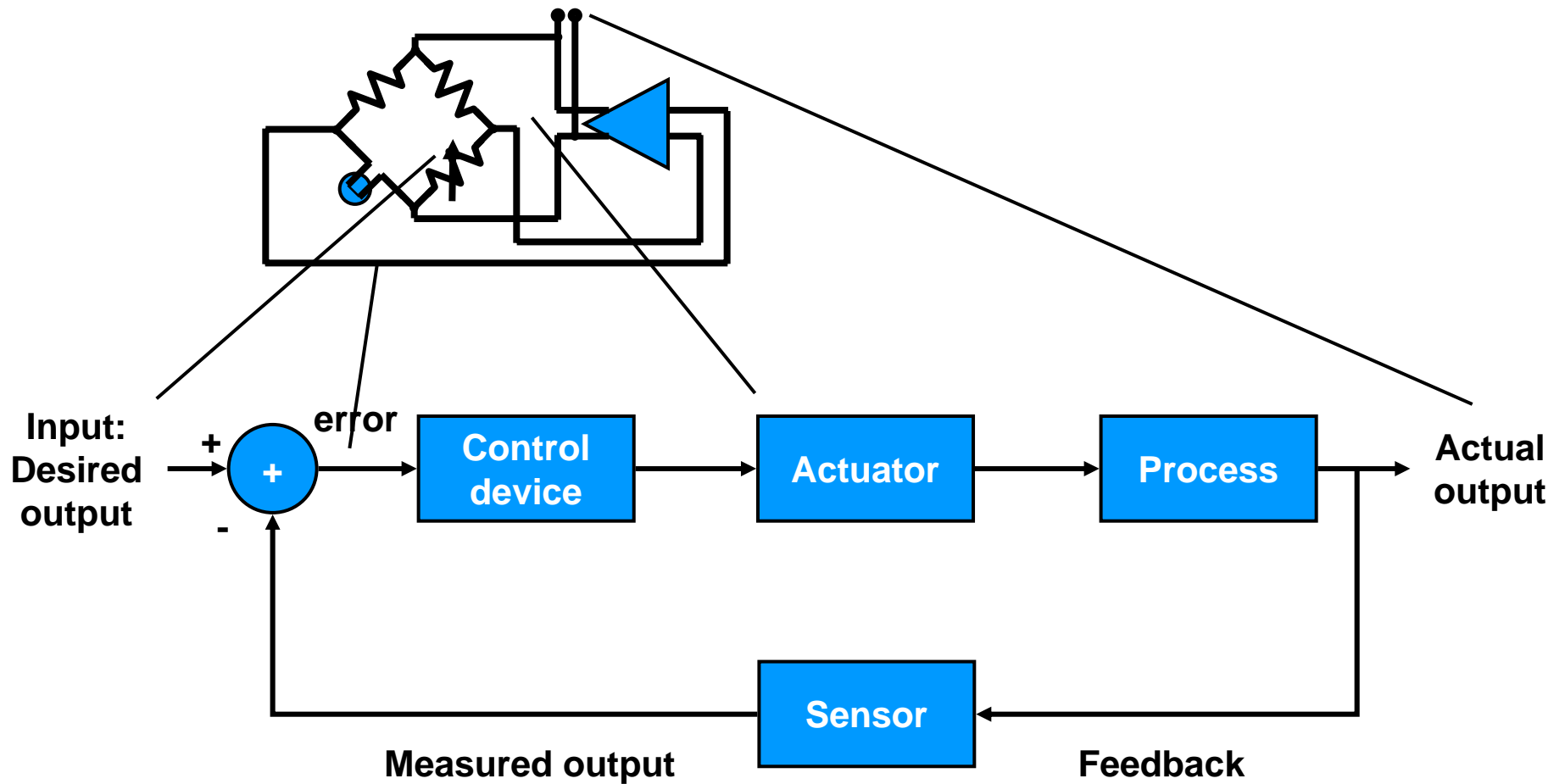
# Negative Feedback Control System



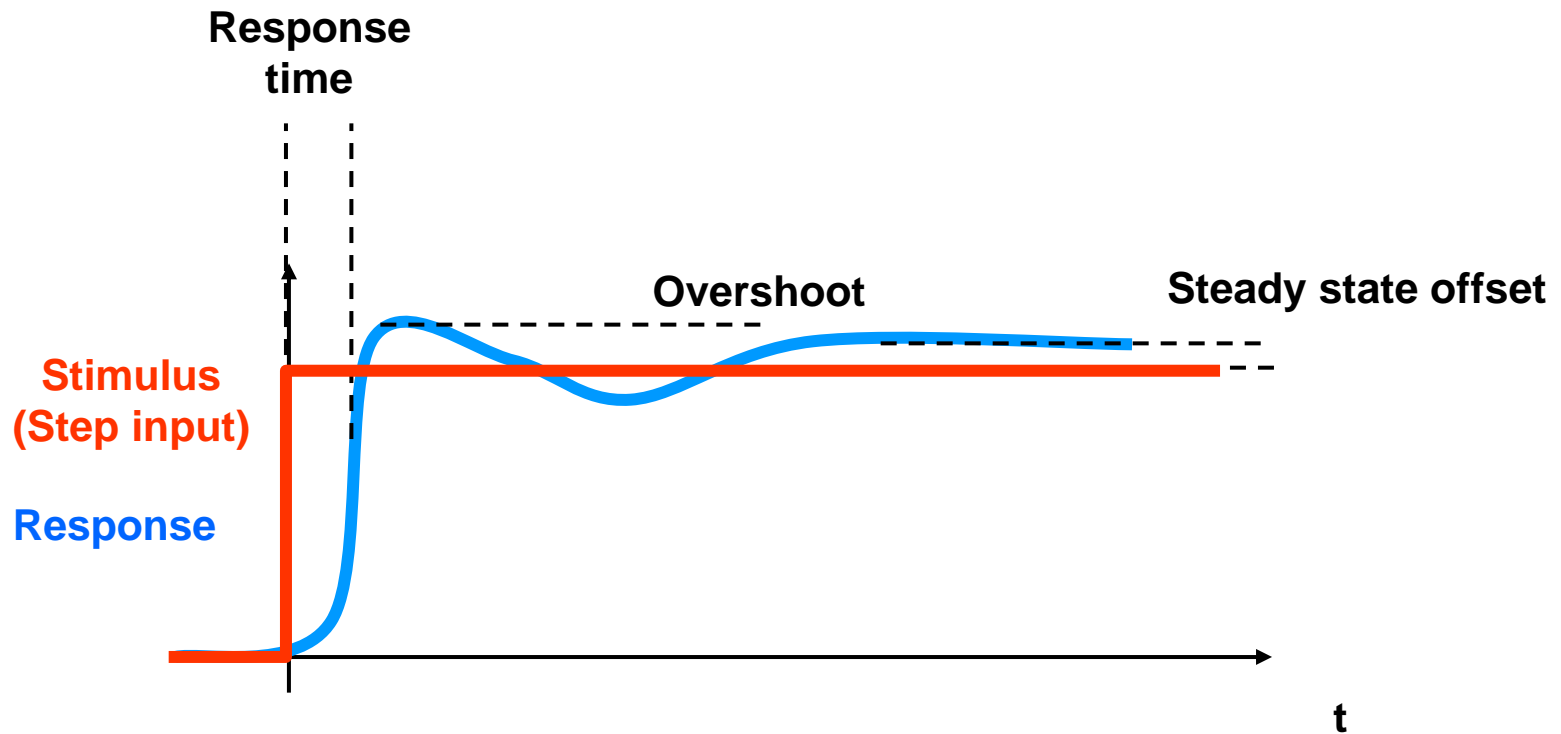
# Application of Hot-wire Anemometer



# Hot-wire Anemometer Feedback Control System



# Issues in Control Systems



# Generalizing The Fourier Series

- Start with the complex Fourier Series

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

# Generalizing The Fourier Series

- Change variables

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

**Replace  $2\pi/T$  with  $\omega_0$**

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jk\omega_0 t} dt$$

# Generalizing The Fourier Series: The Fourier Transform

- Consider what happens when the analysis period is allowed to increase

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi nt}{T}}$$

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

Replace  $2\pi/T$  with  $\omega_0$

Let  $\omega_0$  go to 0  
T becomes infinite

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi kt}{T}} dt$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jk\omega_0 t} dt$$

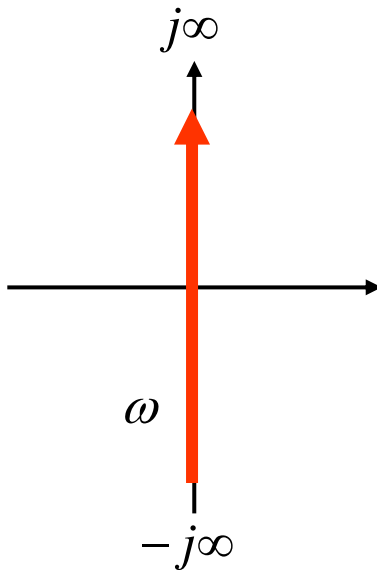
$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

# Generalizing The Fourier Transform

- The Fourier Transform works well with sinusoidal and oscillatory signals
- The Fourier Integral inherently assumes the signal lies somewhere on the  $j\omega$  axis

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

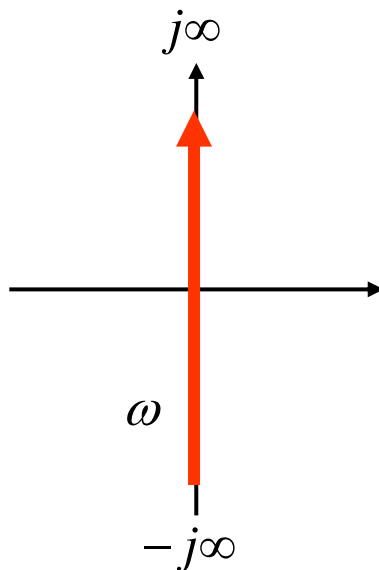


# Generalizing The Fourier Transform

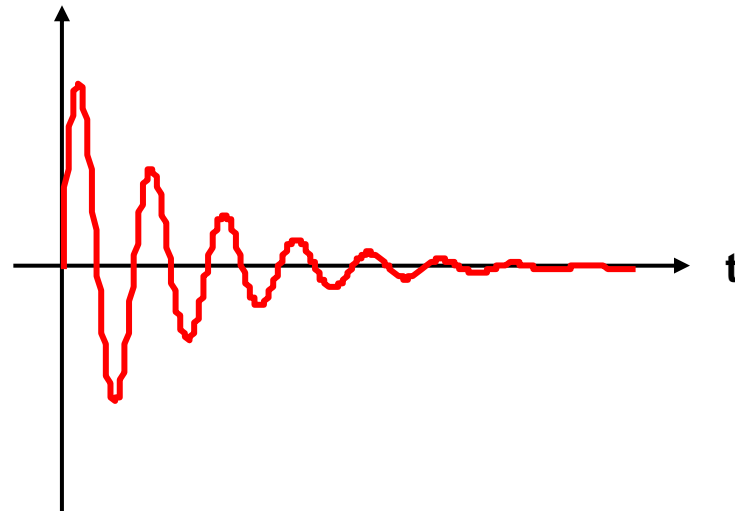
- The Fourier Transform works well with sinusoidal and oscillatory signals
- The Fourier Integral inherently assumes the signal lies somewhere on the  $j\omega$  axis

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$



- But signals in control systems generally exhibit damped or decaying behavior, which the Fourier Transform cannot readily represent

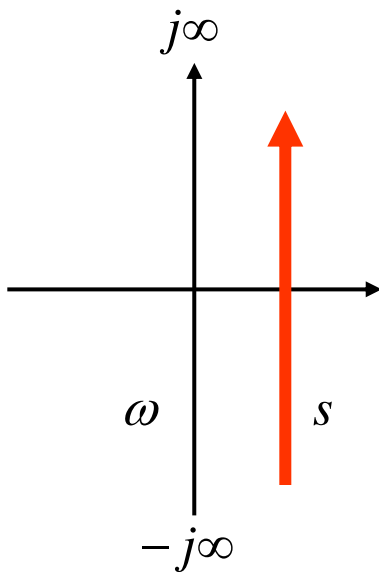


# Generalizing The Fourier Transform: The Laplace Transform

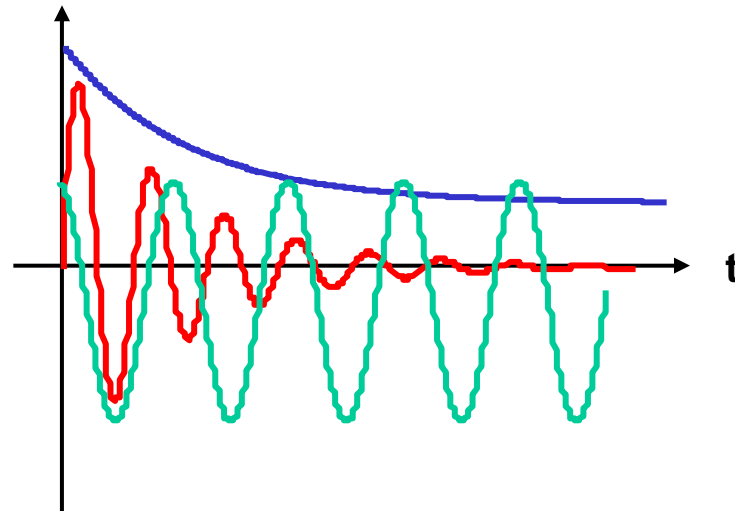
- The Laplace Transform is a generalization of the Fourier Transform with a transform operator that represents oscillatory as well as decaying oscillations

$$f(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} F(s)e^{st} ds$$

$$F(s) = \int_{0-}^{\infty} f(t)e^{-st} dt$$



- The Laplace Transform can deal with a wider variety of signals than the Fourier Transform can.

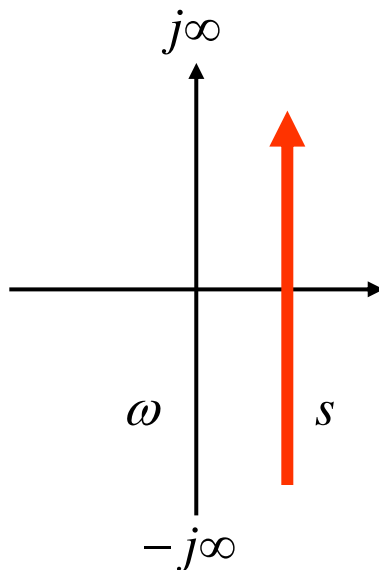


# Generalizing The Fourier Transform: The Laplace Transform

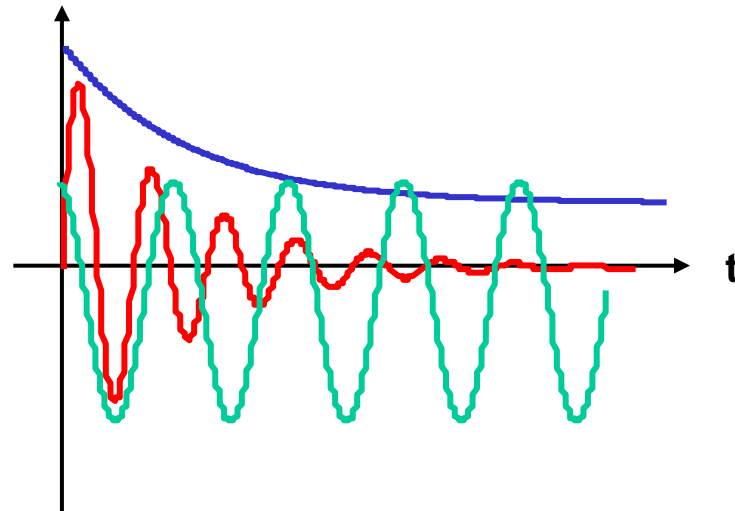
- The Laplace Transform is a generalization of the Fourier Transform with a transform operator that represents oscillatory as well as decaying oscillations

$$f(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} F(s)e^{st} ds$$

$$F(s) = \int_{0-}^{\infty} f(t)e^{-st} dt$$



- The Laplace Transform can deal with a wider variety of signals than the Fourier Transform can.



- The Laplace Transform provides a straightforward way to transform differential equations into algebraic equations, which can be more easily solved.

# Homework 5

- Problems 5.2, 5.8 & 5.10 (use Excel, Matlab, Mathcad, or Scientific Notebook, as you prefer)